Saitek

Technical description for 2.4GHz Wireless Game Pad

LEVEL 1. Rev03

Project Code / Article Number	Product Description	Product Name
PP19	**RF PC Game pad**	**P2900**

Revision History				
		Approved		Change Description
		Engineering	Quality	
Issue Status	Issue Date			
Level 1 (Draft)	**25/01/05**			
Level 1	**04/04/05**			**Updated Project codes, updates in Red**
Level 2 (Draft)				
Level 2				

1.0 General Products Outline

The 2005 Gamepad proposal is to be a derivative of the existing pad design. Enhancements will be functional and aesthetic, with close attention paid to the cost of goods, which must be met in order to achieve such a competitive price point.

Game play enhancements will include a 'free-look' feature, activated by a single button press. This loads a default controller setting for first person shooter games. This is required because most PC games in this genre only allow for mouse and keyboard configuration in game; currently if the gamer wishes to use a pad for fps games they must create a profile, which is time consuming and not very user friendly considering the core target

It is important that the changes made to the new pad derivatives, are subtle enough to please existing customers and different enough to guarantee listings as a new product.

The Pad should also have a solid robust feel, stand out from the main competitors, and provide the user with no lag, no jam, no cable clutter, and no inconvenience.

1.3 Product features PP19 Wireless Pad

- Fantastic looks and design

- Saitek branded logo.
- Chrome effect, Stripe around Shield area
- FPS Button (First person Shooter) Backlit
- Analog / Digital switch
- Dual Shoulder buttons
- USB 2.0 Compatible, true plug and play, and easy to use
- Attractive Packaging that compliments the Pad Design, and act as true Point of Sale merchandise
- Designed and based around the existing P3000 Wireless pad.
- Age Range Usage: 15 to 35 years (mainly male)
- Average Gaming Time: 3 to 7 days a week – 5 to 10hrs Maximum.
- Types of Game Genre: FPS and Sports
- RF design to achieve at least 50 hours continuous play on one AA battery
- User friendly software installation.
- Product tutorial software
- Installation of full colour manuals
- In box "quick start guide"
- SD5 drivers with "Euston" GUI
- Main competitor(s); Logitech

1.4 Industrial Design Considerations

- Overall styling based on the Current P3000 but will require complete new pad Family tooling.
- Component count to be minimal
- Ergonomic Design, with Comfortable grip and feature set. Full head and hand rest adjustability and the Positioning, Spacing, and Feeling of Buttons is also very important.
- Battery life indicator lit through shield area.
- Product Cog is critical to the success of this product.
- Product to offer a 2-year warranty
- RF Dongle to be small and non interfering within the USB hub / port environment similar to PS30
- RF Dongle storage facility on the Pad would be desired

2.0 Product Functional Features.

Features Software	**PP19**
Protocol	USB only connection, Version 2.0 Compliant. USB Certification Required via USB Plug fest.
USB Identification	Vendor ID: 0x06A3 Product ID: 0x040C

Drivers and Programmability	HID Compatible. Fully Programmable. Saitek Driver SD5 GUI, Euston CD Supplied. Product tutorial software
Product ID Script Descriptor	Saitek P2900 Wireless Pad
Architecture Compatibility	• Windows 2000. • Windows XP 32 & 64BIT. • MAC. (OS9 onwards) • Linux. (Kernel 2.4.5 AC9) • Direct X Ver.8A or Later. • HID / PID
Calibration	Auto calibrated; This product should be ready to go when plugged into the users PC, without the need of any physical adjustment, i.e. Without movement of the sticks into all corners by the user.
64 BIT support	Yes
Languages supported on release	English French German Italian Spanish Chinese TBD
Analog / Digital switch	Yes
Features Power	**PP19**
Power Source	1x AA Energiser Battery
Battery Life / Playing time	Alkaline 50 Hours minimum non stop play Re-Chargeable 50 Hours Minimum non stop play. Battery life based on 3,135 mAh
Battery Low Prompt and Indicator	YES via LED

Features Function	**PP19**
Axes	4 Analogue Axes X and Y on left Thumb-stick. Z Axis and Z Rotation on Right Thumb-stick
Analog Control Sticks	2
Internal Analog Stick Button	Yes
Main Buttons	6 (Membrane) on pad face
Shoulder Buttons	4 (Membrane) = 2 on each Shoulder
D-Pad (POV)	Yes: (8 Way) on Head 4 (Tact Switch)
Throttle	NO
Scan Button	YES; for noisy environments and for manual channel hopping when multiple 2.4Ghz devices are present in the same surroundings *Location on shield*
On / Off Switch	YES
Mode select Rumble On / Off	NO
Rumble On / Off LED	NO
L.C.D. Display	NO
Rumble	NO
Range	Up to 10 meters, no less than 9.5M
Cable length	No Cable
Features	**PP19**
Saitek Branding	Yes
Packaging Style	New Branding 2005-6
Warranty	2 Year Warranty (Product Components to be Specified to this)

USP	FPS button (loads "First person shooter" Profile when Programming software is installed) Backlit

3.0 PP19 Product Specific Detail

3.1 Theory of Operation

PP19 is a USB Only RF Game-Pad, with 18 Buttons (including D-Pad), and 4 Axes.

The MCU reads the Pad Input Data and then transmits this to the Receiver Module, via RF transmissions.

3.2 Software and Drivers

Driver Software: Should be based on the standard USB HID Generic Driver as this is a USB only device. Drivers to be digitally signed (i.e. WHQL approved) and also compatible in pre Windows XP OS's.

The PP19 retails with a fully programmable software package, supplied on a CD.

The CD will also have interactive tutorial and full printable colour manuals.

This software will be available in Beta format during development and final product ready for Masproduction. The software and Install guide will be available in the following languages

a) English

a) Spanish

b) French

c) Italian

d) German

e) Chinese TBD

The driver will be integrated into the latest Saitek software build (EUSTON) from 1st Mass Production.

Note: ***Software tested to DP-ENG-008***
WHQL tested to SQS-D-006
USB 2.0 tested to SQS-D-005

3.3 Installation process

Once the CD is inserted into the user's machine, the following options will be shown on the Saitek applet.

Exact text TBD

1) View Manuals; Two choices:- a) Installation Manual b) SST Programming Manual
2) Install Drivers and SST Programming Software
3) Product Tour
4) View Product Catalogue
5) Visit The Saitek Website
6) Exit

Option 1 provides the user with the choice to view the Installation manual and the SST Programming manual. Also there is an option to install both manuals to the user's hard drive.

Option 2 Proceeds with installing the product drivers and then offering the user to option to install the SST Programming Software.

Option 3 This is an interactive Tutorial to help the user become more familiar with the features of the PS30, it will take the form of a 3D model of the stick with "Hot Spots" that can be clicked on with the mouse and a "Pop Up" will appear and provide information on the particular selected feature.

Option 4 this is a clickable link which will automatically launch the user's browser to the Saitek website.

Option 5 Exits the Saitek installation process or closes the applet after installation has been completed.

3.4 Registration Process

After the software install is complete, the user has an option to register their device. This requires clicking the Check box in the Registration screen.

If the user chooses not to register their device, they can click finish at the bottom of the registration screen and the finish screen is displayed.

3.5 Architecture

Driver Software: As driver section.

Compatibility Requirements:-

- USB Compliant to Version 2.0
- WHQL
- Windows 2000, Windows XP 32 BIT, Windows XP 64 BIT and Mac.
- Direct X Ver.6 or Later.
- HID / PID specification for Device Input and Output.

3.6 Main Processors

MCU:

- DIE26643
- CYWUSB6934

4.0 Electronic Operation

4.1 Electronics design brief

The Electronics will be designed and developed with production, and pricing category, in mind. Thus a low cost Electronics solution meeting pricing targets should be employed.

USB: The Game-Pad should be fully USB2.0 compatible, and subsequently USB1.1 Backwards compatible.

EMC / FCC: All Electronics components selected for Design, including PCB's / Discrete / Semiconductors should be fully EMC / FCC compatible. Subsequently the total Electronics solution will be designed with EMC/FCC in mind at ALL stages of development. This will be backed up with test results, pre-scan data and approval during development. The stick should be EMC / FCC approved before Maspro starts, and ideally before Pilot Production.

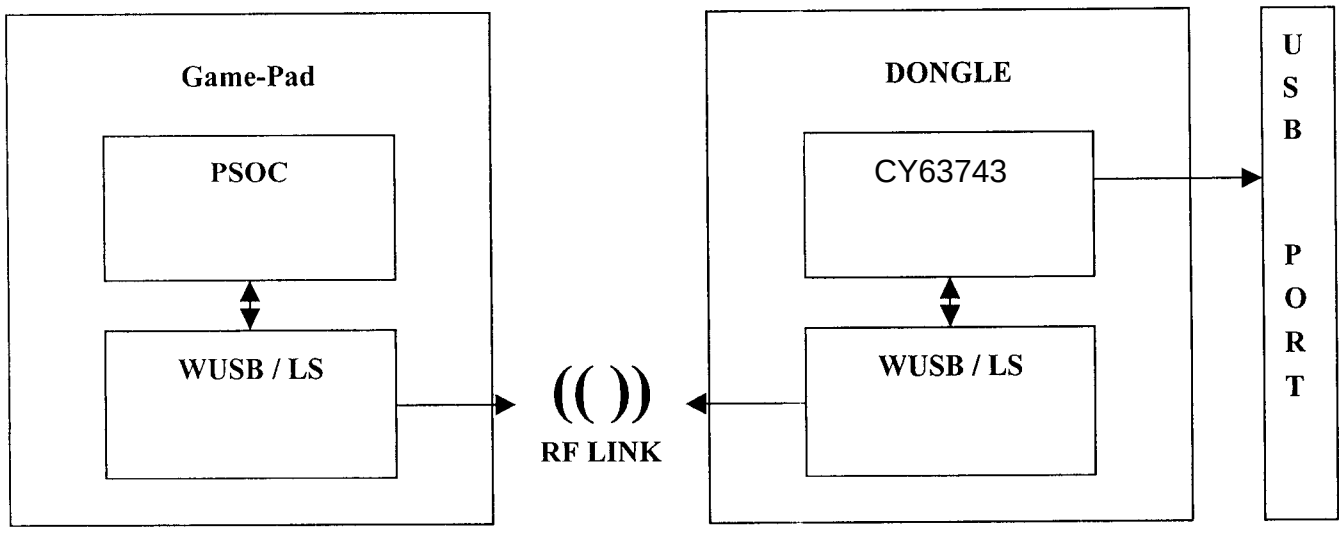
Other tests that will be required are as follows;

1) LVD testing.
2) For Europe - R&TTE compliance report and certificate.
3) For Canada 210 markings and statement

All Issued Schematics and PCB / Electronics evaluation samples, either Inputted or Outputted by all concerning parties, e.g., SP and SEE, should be fully checked and approved by the Designer / Electronics Manager / Project Manager, and VO. Please follow procedure; **DP-ENG-004**

4.2 Diagram of operation

The Electronics solution requires the following to operate;



4.3 PSoC

Operating Frequency	93.7Khz – 24Mhz
Operating Voltage	3.0 – 5.25V
Program Memory	16
Data Memory	256
Digital PSoC Blocks	8
Analogue PSoC Blocks	12
I/O Pins	40/44
External Switch Mode Pump	Yes
Available Packages	48PDIP, 48SSOP, 44TQFP

For full specification please refer to document; CY8C26443-24PSOC.PDF

4.4 WUSB

4.4.1 Features

- 2.4-GHz radio transceiver
- Operates in the unlicensed Industrial, Scientific, and Medical (ISM) band (2.4-GHz–2.483GHz)
- -90dBm receive sensitivity
- Up to 0dBm output power
- Range of up to 10 meters or more
- Data throughput of up to 62.5kbits/sec

- Highly integrated low cost, minimal number of external components required
- Dual DSSS reconfigurable baseband correlators
- SPI microcontroller interface (up to 2-MHz data rate)
- 13 MHz ± 50-ppm input clock operation
- Low standby current ~ 1 µA
- Integrated 32 bit Manufacturing ID
- Operating voltage from 2.7V to 3.6V
- Operating temperature from 0° to 70°C
- Offered in a small footprint 48 Quad Flat Pack No Leads (QFN) or cost saving 28-lead exposed paddle SOIC

4.4.2 Wireless USB LS System Overview.

WirelessUSB LS adds to the existing WirelessUSB portfolio a low-cost 2.4-GHz wireless solution that enables the wireless gaming console peripheral device market and displaces the 27-MHz solutions currently used for low-end retail PC Human Interface Device (HID) applications. A WirelessUSB LS system typically consists of a WirelessUSB LS Bridge and at least one WirelessUSB LS HID. The host PC is not aware of the wireless connection, since the interface to the host acts like a normal wired USB HID connection. Therefore, there is no special software required on the host PC in order to support WirelessUSB LS.

For full specification please refer to document; CYWUSB6934.PDF

4.5 CY63743

Operating Voltage	4.0 – 5.5VDC
Internal Memory	256MB, 8Kbytes EEPROM
I/O Ports	Up to 16
Internal Oscillator	Yes 12Mhz
Available Packages	24PDIP, 24SOP

For full specification please refer to document; CY7C63743.PDF

5.0 RF Design Specification and Functional Overview

5.1 Theory of Operation

The Game-Pad and the Dongle will power up and connect using channel information stored in the EEPROM. This is ensures that a known good channel is used, and minimises risk of interference. Data communications will remain on this channel, until the user selects a different channel.
The user can select a different channel by pressing the "Scan" button on the Game-Pad. The Game-Pad will then perform a channel negotiation routine with the Dongle, and a new channel will be selected. No user intervention is required on the Dongle

5.2 Transmit mode

The PP19 is an RF Game-Pad. It will consist of 2 units: a Dongle, which connects to the PC, and facilitates USB data transfer; and a remote Game-Pad which controls RF data transfer to the Dongle. All data transfer will be initialised by the Game-Pad, and will only occur, when an update of the data is required.
The Dongle will respond with an acknowledge transmission, such that the Game-Pad can ensure that the data has successfully been transferred to the Dongle.

All RF transmissions will be using Direct Sequence Spread Spectrum (DSSS) modulation, which will meet FCC and ETSI emission regulations.

5.2.1 Game-Pad processor

PSoC; DIE26643

5.2.2 Dongle module

The processor will enable the Dongle to enumerate as a Low speed USB peripheral device. The processor will also control an RF transceiver, which enables the Dongle to obtain RF data sent by the Game-Pad, and pass it to the USB Host. An EEPROM is also used to store channel information. The Dongle is powered by the USB 5V supply line, and will draw less than 100mA from the USB line. This will enable the Dongle to be connected to a bus Powered Hub.

5.2.3 Dongle Processor

CY63743

5.2.4 Range

Up to 10 Meters, No less than 9.5M

5.2.5 Antennas

PCB mounted on the RF boards of the Game-Pad and Dongle.

5.2.6 Power supplies

Pad = 1 AA battery 1.5V and Re-Chargeable 1.2V

Dongle = USB port 4.70 to 5.25V, 100mA.

5.2.7 Battery requirements / Warning labels.

Please follow procedures:

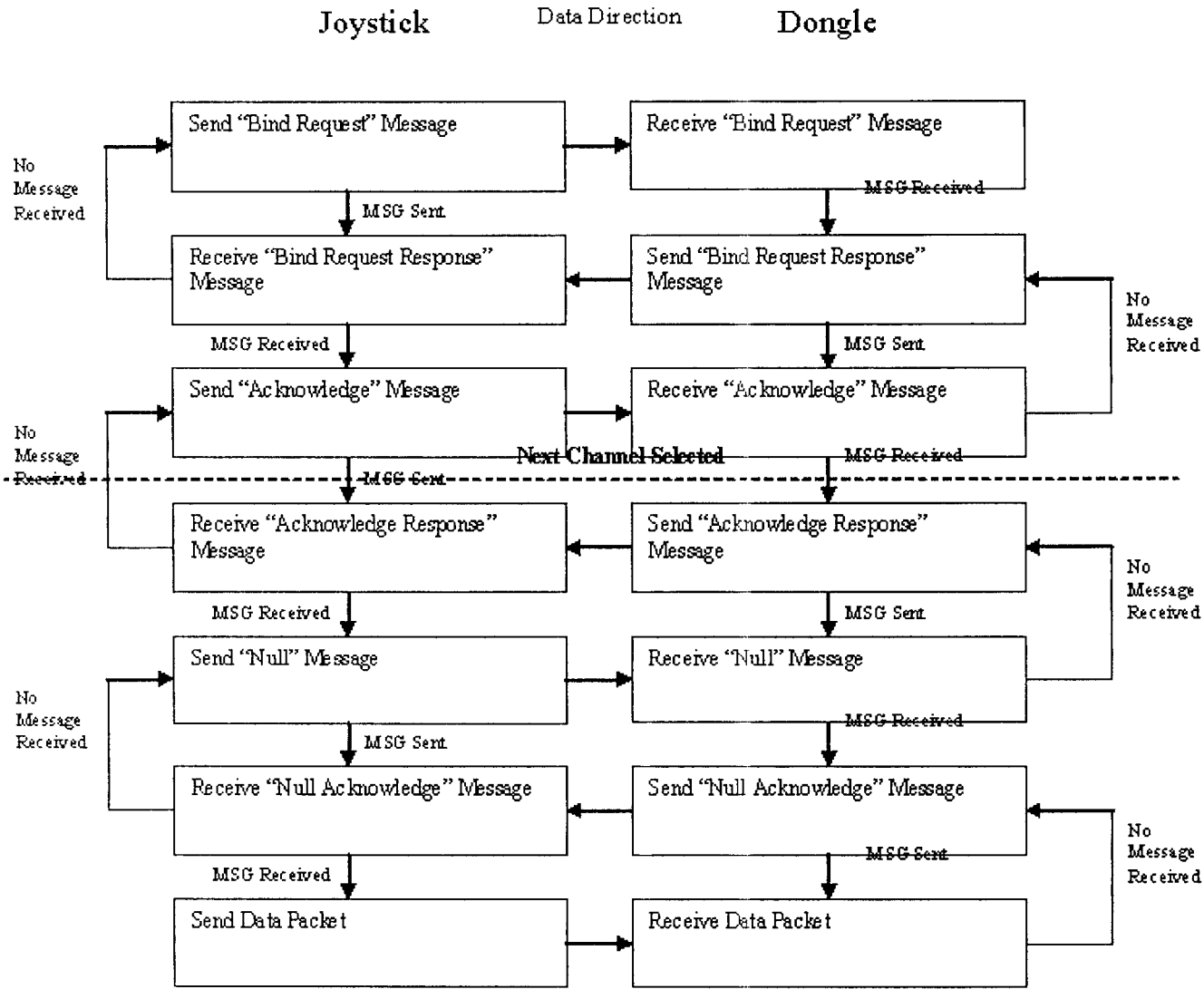
Battery contents and requirements to SQS-S-012

Battery compartment design requirements to SQS-S-013

5.2.8 Power management

TBD

5.2.9 Multiplayer and Channel select theory of Operation



Next Channel Negotiation Process

6.0 Button Key Assignments for PP19

Feature	Direct X Button No.	Position
Control Button	1	Top Face Right
Control Button	2	Top Face Right
Control Button	3	Top Face Right
Control Button	4	Top Face Right
Shoulder Button L1	5	Top Side Left
Shoulder Button R1	6	Top Side Left
Shoulder Button L2	7	Top Side Right
Shoulder Button R2	8	Top Side Right
Function Button 1	9	Top Face Right

Function Button 2	10	Top Face Right
Mini Stick Button (Left)	11	Underneath Joystick Left
Mini Stick Button (Right)	12	Underneath Joystick Right
FPS Button, Backlit	13	Top Face Right
Analog Digital Button	14	Shield Top Face Center
ON / OFF Switch	NA Hard Wired	Underside
Scan Button	NA	Shield
D-Pad	NA	Top Face Left

7.0 Mechanical Actuation Forces

Switch Type		
D-Pad	**0.2 ~ 0.4 N**	**Measured at End Tips of D-pad**
Control Buttons	**0.1 ~ 0.2 N**	**Measured at centre of switch**
Triggers	**0.1 ~ 0.2 N**	**Measured at centre of switch**
Other Buttons	**0.1 ~ 0.2 N**	**Measured at centre of switch**

8.0 Statutory Compliance tests

8.1 Maximum Applied Loadings.

Note: All tests are for 10% of product rated life

Component / Feature	**Maximum load**	**Applied**
Trigger	50N for 15 Seconds	Centre of Switch
D-pad	50N for 15 Seconds	End point of D-Pad
Control and Analogue Mini-Stick Buttons	50N for 15 Seconds	Centre of Button
Other Buttons	30N for 15 Seconds	Centre of Button
Cable pull force (USB only)	100N for 15 seconds	

9.0 Environmental (IMPORTANT) and Raw Material Marking

All plastic components where possible must be permanently identified with re-cycle information.

Large Mouldings to include, Mould Clocks indication; Shift/day/month/year of moulding.

All packaging to be in bio-degradable materials

Note: All markings to be invisible to USER

10.0 Packaging Requirements.

Style and Packaging Type to be confirmed but is tentatively as follows. Please confirm with Product and Project Manager before proceeding.

The product is packaged in a cardboard gift box.

Gift box requirement:

- 2.0mm E-flute cardboard with 260gms single Art paper
- 4-colour printing
- Calendar finishing
- CE mark(print)
- FCC mark(print)
- Canada Mark (print)
- Article No.(print)
- IMAC Icon(print)
- Recycle mark(print)
- USB Icon(print)
- Barcode (print)
- Windows XP logo (Print)

The external gift-box dimensions must be kept to an absolute minimum and fully approved by SP MSD & SP Eng. prior to release for manufacture.

Each gift-box shall be multi-packed into an export carton. The external dimensions for the export carton must comply with Saitek Standard PD-QA-04-006 v1- Export carton Design & Marking.

11.0 Service Manuals / Repair Requirements.

The product range will require a fully intuitive repair /service manual that clearly indicates serviceable components and methodology for repair or replacement, as per PD-QA-05-002 v2

The design will be modular in that all serviceable components (if any) can be disassembled and replaced with minimum labour / skill.

Frequency Table

Item	Frequency (MHz)	Item	Frequency (MHz)
1	2408	35	2448
2	2409	36	2449
3	2410	37	2450
4	2411	38	2451
5	2412	39	2452
6	2413	40	2453
7	2416	41	2454
8	2417	42	2455
9	2418	43	2456
10	2419	44	2457
11	2420	45	2458
12	2421	46	2459
13	2422	47	2460
14	2423	48	2461
15	2424	49	2464
16	2425	50	2465
17	2426	51	2466
18	2427	52	2467
19	2428	53	2468
20	2429	54	2469
21	2432	55	2470
22	2433	56	2471
23	2434	57	2472
24	2435	58	2473
25	2436	59	2474
26	2437		
27	2438		
28	2439		
29	2440		
30	2441		
31	2442		
32	2443		
33	2444		
34	2445		

WirelessUSB™ LS 2.4-GHz DSSS Radio SoC

1.0 Features

- **2.4-GHz radio transceiver**
- **Operates in the unlicensed Industrial, Scientific, and Medical (ISM) band (2.4 GHz–2.483 GHz)**
- **-90-dBm receive sensitivity**
- **Up to 0 dBm output power**
- **Range of up to 10 meters or more**
- **Data throughput of up to 62.5 kbits/sec**
- **Highly integrated low cost, minimal number of external components required**
- **Dual DSSS reconfigurable baseband correlators**
- **SPI microcontroller interface (up to 2-MHz data rate)**
- **13-MHz ± 50-ppm input clock operation**
- **Low standby current < 1 µA**
- **Integrated 30-bit Manufacturing ID**
- **Operating voltage from 2.7V to 3.6V**
- **Operating temperature from 0° to 70°C**
- **Offered in a small footprint 48 Quad Flat Pack No Leads (QFN)**

2.0 Functional Description

The CYWUSB6932/CYWUSB6934 Integrated Circuits (ICs) are highly integrated 2.4-GHz Direct Sequence Spread Spectrum (DSSS) Radio System-on-Chip (SoC) ICs. From the Serial Peripheral Interface (SPI) to the antenna, these ICs are single-chip 2.4-GHz DSSS Gaussian Frequency Shift Keying (GFSK) baseband modems that connect directly to a micro-controller via simple serial interface.

The CYWUSB6932 transmit-only IC and the CYWUSB6934 transceiver IC are available in a small footprint 48-pin QFN package.

3.0 Applications

- PC Human Interface Devices (HIDs)
  - Mice
  - Keyboards
  - Joysticks
- Peripheral Gaming Devices
  - Game Controllers
  - Console Keyboards
- General
  - Presenter Tools
  - Remote Controls
  - Consumer Electronics
  - Barcode Scanners
  - POS Peripherals
  - Toys

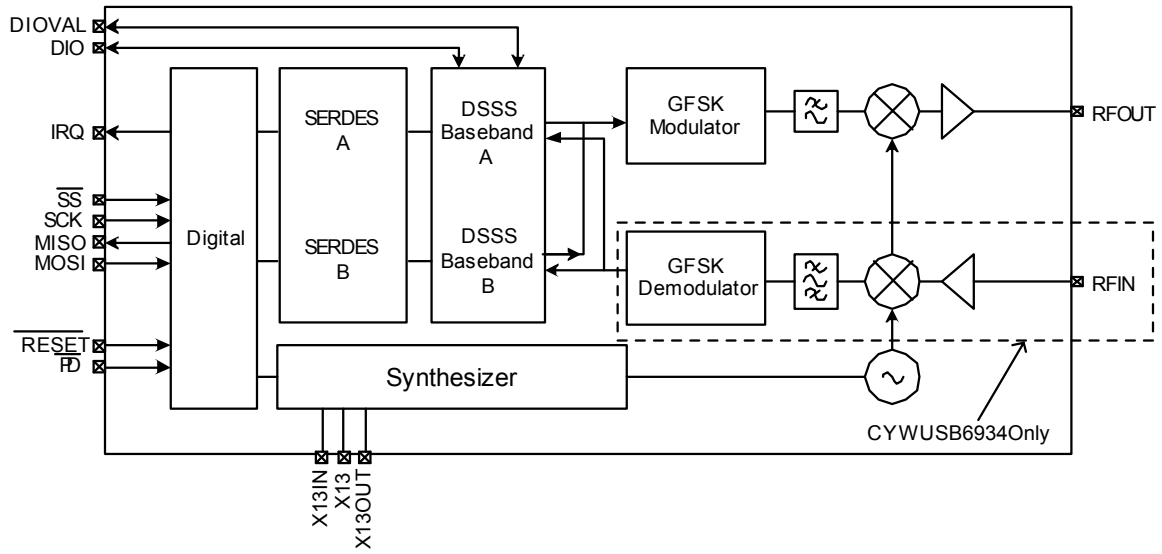


Figure 3-1. CYWUSB6932/CYWUSB6934 Simplified Block Diagram

3.1 Applications Support

The CYWUSB6932/CYWUSB6934 ICs are supported by the CY3632 WirelessUSB Development Kit. The development kit provides all of the materials and documents needed to cut the cord on wired applications including two radio modules that connect directly to two prototyping platform boards, comprehensive WirelessUSB protocol code examples a WirelessUSB Listener tool and all of the associated schematics, gerber files and bill of materials.

The CY4632 WirelessUSB LS Keyboard Mouse Reference Design provides a production-worthy example of a wireless mouse and keyboard system.

The CY3633 WirelessUSB LS Gaming Development Kit provides support for designing a wireless gamepad for the major gaming consoles and is offered as an accessory to the CY3632 WirelessUSB.

4.0 Functional Overview

The CYWUSB6932/CYWUSB6934 ICs provide a complete WirelessUSB LS SPI to antenna radio modem. The SoC is designed to implement wireless devices operating in the worldwide 2.4-GHz Industrial, Scientific, and Medical (ISM) frequency band (2.400 GHz–2.4835 GHz). It is intended for systems compliant with world-wide regulations covered by ETSI EN 301 489-1 V1.4.1, ETSI EN 300 328-1 V1.3.1 (European Countries); FCC CFR 47 Part 15 (USA and Industry Canada) and ARIB STD-T66 (Japan).

The CYWUSB6934 IC contains a 2.4-GHz radio transceiver, a GFSK modem and a dual DSSS reconfigurable baseband. The CYWUSB6932 IC contains a 2.4-GHz radio transmit-only, a GFSK modem and a DSSS baseband. The radio and baseband are both code- and frequency-agile. Forty-nine spreading codes selected for optimal performance (Gold codes) are supported across 78 1-MHz channels yielding a theoretical spectral capacity of 3822 channels. Both ICs support a range of up to 10 meters or more.

4.1 2.4-GHz Radio

The receiver and transmitter are a single-conversion low-Intermediate Frequency (low-IF) architecture with fully integrated IF channel matched filters to achieve high performance in the presence of interference. An integrated Power Amplifier (PA) provides an output power control range of 30 dB in seven steps.

Table 4-1. Internal PA Output Power Step Table

PA Setting	Typical Output Power (dBm)
7	0
6	–2.4
5	–5.6
4	–9.7
3	–16.4
2	–20.8
1	–24.8
0	–29.0

Both the receiver and transmitter integrated Voltage Controlled Oscillator (VCO) and synthesizer have the agility to cover the complete 2.4-GHz GFSK radio transmitter ISM band. The synthesizer provides the frequency-hopping local oscillator for the transmitter and receiver. The VCO loop filter is also integrated on-chip.

4.2 GFSK Modem

The transmitter uses a DSP-based vector modulator to convert the 1-MHz chips to an accurate GFSK carrier.

The receiver uses a fully integrated Frequency Modulator (FM) detector with automatic data slicer to demodulate the GFSK signal.

4.3 Dual DSSS Baseband

Data is converted to DSSS chips by a digital spreader. De-spreading is performed by an oversampled correlator. The DSSS baseband cancels spurious noise and assembles properly correlated data bytes.

The DSSS baseband has three operating modes: 64 chips/bit Single Channel, 32 chips/bit Single Channel, and 32 chips/bit Single Channel Dual Data Rate (DDR).

4.3.1 64 Chips/Bit Single Channel

The baseband supports a single data stream operating at 15.625 kbits/sec. The advantage of selecting this mode is its ability to tolerate a noisy environment. This is because the 15.625 kbits/sec data stream utilizes the longest PN Code resulting in the highest probability for recovering packets over the air. This mode can also be selected for systems requiring data transmissions over longer ranges.

4.3.2 32 Chips/Bit Single Channel

The baseband supports a single data stream operating at 31.25 kbits/sec.

4.3.3 32 Chips/Bit Single Channel Dual Data Rate (DDR)

The baseband spreads bits in pairs and supports a single data stream operating at 62.5 kbits/sec.

4.4 Serializer/Deserializer (SERDES)

The CYWUSB6934 IC has a data Serializer/Deserializer (SERDES), which provides byte-level framing of transmit and receive data. Bytes for transmission are loaded into the SERDES and receive bytes are read from the SERDES via the SPI interface. The SERDES provides double buffering of transmit and receive data. While one byte is being transmitted by the radio the next byte can be written to the SERDES data register insuring there are no breaks in transmitted data.

After a receive byte has been received it is loaded into the SERDES data register and can be read at any time until the next byte is received, at which time the old contents of the SERDES data register will be overwritten. The CYWUSB6932 IC only has a data Serializer.

4.5 Application Interfaces

Both ICs have a fully synchronous SPI slave interface for connectivity to the application MCU. Configuration and byte-oriented data transfer can be performed over this interface. An interrupt is provided to trigger real time events.

An optional SERDES Bypass mode (DIO) is provided for applications that require a synchronous serial bit-oriented data path. This interface is for data only.

4.6 Clocking and Power Management

A 13-MHz crystal (±50 ppm or better) is directly connected to X13IN and X13 without the need for external capacitors. Both ICs have a programmable trim capability for adjusting the on-chip load capacitance supplied to the crystal. The Radio Frequency (RF) circuitry has on-chip decoupling capacitors. Both devices are powered from a 2.7V to 3.6V DC supply. Both devices can be shutdown to a fully static state using the $\overline{PD}$ pin.

Below are the requirements for the crystal to be directly connected to X13IN and X13:

- Nominal Frequency: 13 MHz
- Operating Mode: Fundamental Mode
- Resonance Mode: Parallel Resonant
- Frequency Stability: ± 50 ppm
- Series Resistance: ≤ 100 ohms
- Load Capacitance: 10 pF
- Drive Level: 10 uW–100 uW

4.7 Receive Signal Strength Indicator (RSSI)

The RSSI register (Reg 0x22) (applies only to the CYWUSB6934 IC) returns the relative signal strength of the ON-channel signal power and can be used to:

1. Determine the connection quality
2. Determine the value of the noise floor
3. Check for a quiet channel before transmitting.

The internal RSSI voltage is sampled through a 5-bit analog-to-digital converter (ADC). A state machine controls the conversion process. Under normal conditions, the RSSI state machine initiates a conversion when an ON-channel carrier is detected and remains above the noise floor for over 50 µs. The conversion produces a 5-bit value in the RSSI register (Reg 0x22, bits 4:0) along with a valid bit, RSSI register (Reg 0x22, bit 5). The state machine then remains in HALT mode and does not reset for a new conversion until the receive mode is toggled off and on. Once a connection has been established, the RSSI register can be read to determine the relative connection quality of the channel. A RSSI register value lower than 10 indicates that the received signal strength is low, a value greater than 28 indicates a strong signal level.

To check for a quiet channel before transmitting, first set up receive mode properly and read the RSSI register (Reg 0x22). If the valid bit is zero, then force the Carrier Detect register (Reg 0x2F, bit 7=1) to initiate an ADC conversion. Then, wait greater than 50 µs and read the RSSI register again. Next, clear the Carrier Detect Register (Reg 0x2F, bit 7=0) and turn the receiver OFF. Measuring the noise floor of a quiet channel is inherently a 'noisy' process so, for best results, this procedure should be repeated several times (~20) to compute an average noise floor level. A RSSI register value of 0-10 indicates a channel that is relatively quiet. A RSSI register value greater than 10 indicates the channel is probably being used. A RSSI register value greater than 28 indicates the presence of a strong signal.

5.0 Application Interfaces

5.1 SPI Interface

The CYWUSB6932/CYWUSB6934 ICs have a four-wire SPI communication interface between an application MCU and one or more slave devices. The SPI interface supports single-byte and multi-byte serial transfers. The four-wire SPI communications interface consists of Master Out-Slave In (MOSI), Master In-Slave Out (MISO), Serial Clock (SCK), and Slave Select ($\overline{SS}$).

The SPI receives SCK from an application MCU on the SCK pin. Data from the application MCU is shifted in on the MOSI pin. Data to the application MCU is shifted out on the MISO pin. The active-low Slave Select ($\overline{SS}$) pin must be asserted to initiate a SPI transfer.

The application MCU can initiate a SPI data transfer via a multi-byte transaction. The first byte is the Command/Address byte, and the following bytes are the data bytes as shown in *Figure 5-1* through *Figure 5-4*. The $\overline{SS}$ signal should not be deasserted between bytes. The SPI communications is as follows:

- Command Direction (bit 7) = “0” Enables SPI read transaction. A “1” enables SPI write transactions.
- Command Increment (bit 6) = “1” Enables SPI auto address increment. When set, the address field automatically increments at the end of each data byte in a burst access, otherwise the same address is accessed.
- Six bits of address.
- Eight bits of data.

The SPI communications interface has a burst mechanism, where the command byte can be followed by as many data bytes as desired. A burst transaction is terminated by deasserting the slave select ($\overline{SS}$ = 1). For burst read transactions, the application MCU must abide by the timing shown in *Figure 12-2*.

The SPI communications interface single read and burst read sequences are shown in *Figure 5-2* and *Figure 5-3*, respectively.

The SPI communications interface single write and burst write sequences are shown in *Figure 5-4* and *Figure 5-5*, respectively.

	Byte 1			Byte 1+N
Bit #	7	6	[5:0]	[7:0]
Bit Name	DIR	INC	Address	Data

Figure 5-1. SPI Transaction Format

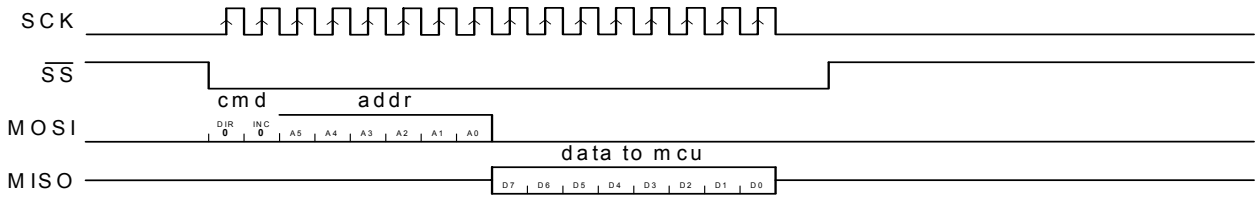


Figure 5-2. SPI Single Read Sequence

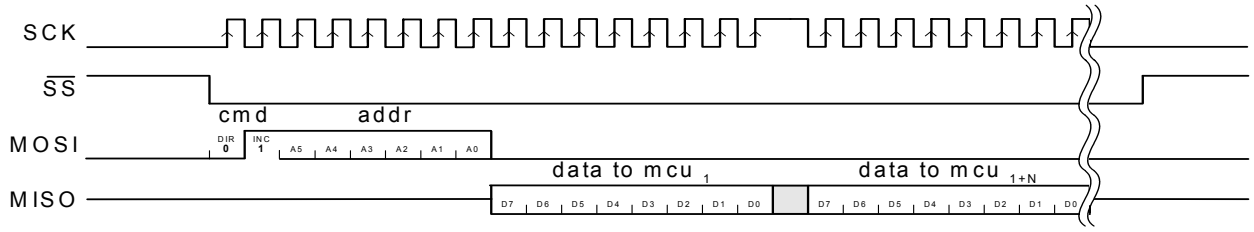


Figure 5-3. SPI Burst Read Sequence

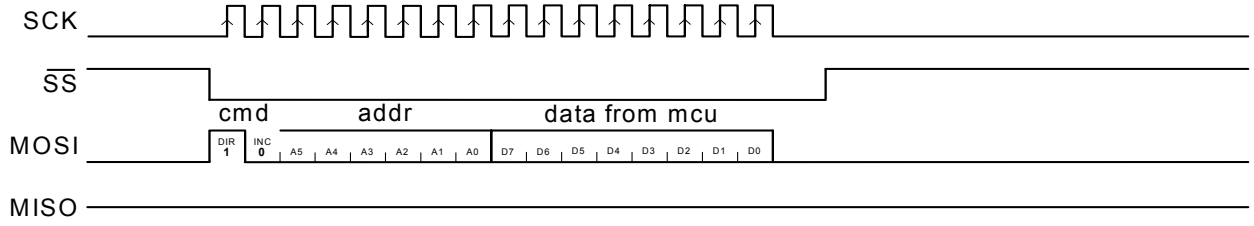


Figure 5-4. SPI Single Write Sequence

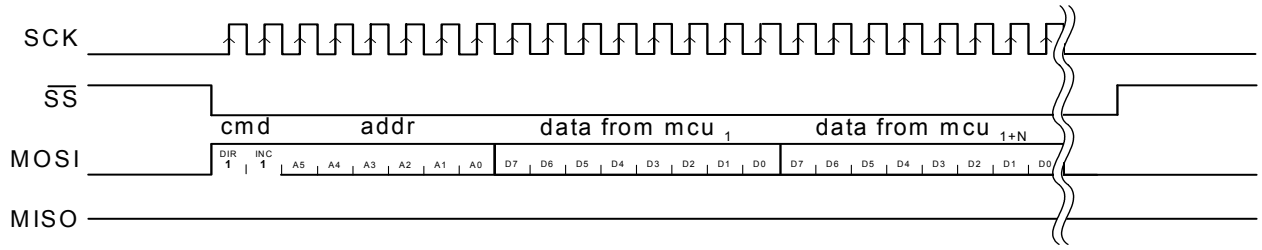


Figure 5-5. SPI Burst Write Sequence

5.2 DIO Interface

The DIO communications interface is an optional SERDES bypass data-only transfer interface. In receive mode, DIO and DIOVAL are valid after the falling edge of IRQ, which clocks the data as shown in *Figure 5-6*. In transmit mode, DIO and DIOVAL are sampled on the falling edge of the IRQ, which clocks the data as shown in *Figure 5-7*. The application MCU samples the DIO and DIOVAL on the rising edge of IRQ.

5.3 Interrupts

The CYWUSB6932/CYWUSB6934 ICs feature three sets of interrupts: transmit, receive (CYWUSB6934 only), and a wake interrupt. These interrupts all share a single pin (IRQ), but can be independently enabled/disabled. In transmit mode, all receive interrupts are automatically disabled, and in receive mode all transmit interrupts are automatically disabled. However, the contents of the enable registers are preserved when switching between transmit and receive modes.

Interrupts are enabled and the status read through 6 registers: Receive Interrupt Enable (Reg 0x07), Receive Interrupt Status (Reg 0x08), Transmit Interrupt Enable (Reg 0x0D), Transmit Interrupt Status (Reg 0x0E), Wake Enable (Reg 0x1C), Wake Status (Reg 0x1D).

If more than 1 interrupt is enabled at any time, it is necessary to read the relevant interrupt status register to determine which event caused the IRQ pin to assert. Even when a given interrupt source is disabled, the status of the condition that would otherwise cause an interrupt can be determined by reading the appropriate interrupt status register. It is therefore possible to use the devices without making use of the IRQ pin at all. Firmware can poll the interrupt status register(s) to wait for an event, rather than using the IRQ pin.

The polarity of all interrupts can be set by writing to the Configuration register (Reg 0x05), and it is possible to configure the IRQ pin to be open drain (if active low) or open source (if active high).

5.3.1 Wake Interrupt

When the $\overline{PD}$ pin is low, the oscillator is stopped. After $\overline{PD}$ is deasserted, the oscillator takes time to start, and until it has done so, it is not safe to use the SPI interface. The wake interrupt indicates that the oscillator has started, and that the device is ready to receive SPI transfers.

The wake interrupt is enabled by setting bit 0 of the Wake Enable register (Reg 0x1C, bit 0=1). Whether or not a wake interrupt is pending is indicated by the state of bit 0 of the Wake Status register (Reg 0x1D, bit 0). Reading the Wake Status register (Reg 0x1D) clears the interrupt.

5.3.2 Transmit Interrupts

Four interrupts are provided to flag the occurrence of transmit events. The interrupts are enabled by writing to the Transmit Interrupt Enable register (Reg 0x0D), and their status may be determined by reading the Transmit Interrupt Status register (Reg 0x0E). If more than 1 interrupt is enabled, it is necessary to read the Transmit Interrupt Status register (Reg 0x0E) to determine which event caused the IRQ pin to assert.

The function and operation of these interrupts are described in detail in *Section 7.0*.

5.3.3 Receive Interrupts

Eight interrupts are provided to flag the occurrence of receive events, four each for SERDES A and B. In 64 chips/bit and 32 chips/bit DDR modes, only the SERDES A interrupts are available, and the SERDES B interrupts will never trigger, even if enabled. The interrupts are enabled by writing to the Receive Interrupt Enable register (Reg 0x07), and their status may be determined by reading the Receive Interrupt Status register (Reg 0x08). If more than one interrupt is enabled, it is necessary to read the Receive Interrupt Status register (Reg 0x08) to determine which event caused the IRQ pin to assert.

The function and operation of these interrupts are described in detail in *Section 7.0*.

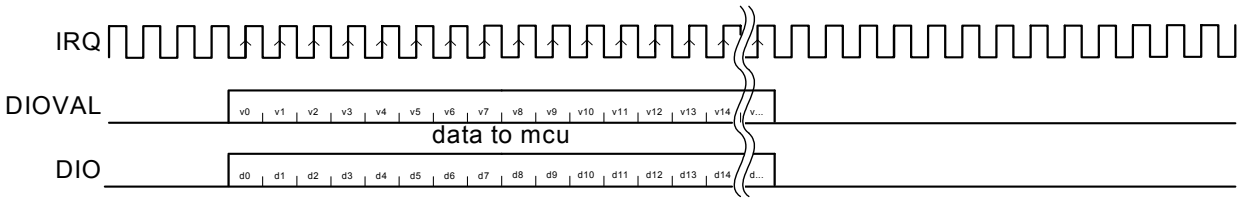


Figure 5-6. DIO Receive Sequence

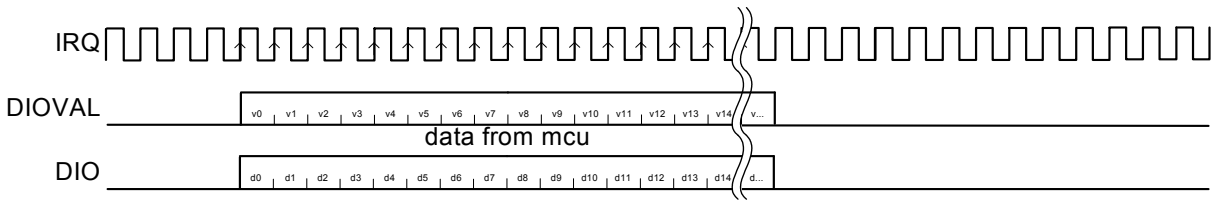


Figure 5-7. DIO Transmit Sequence

6.0 Application Examples

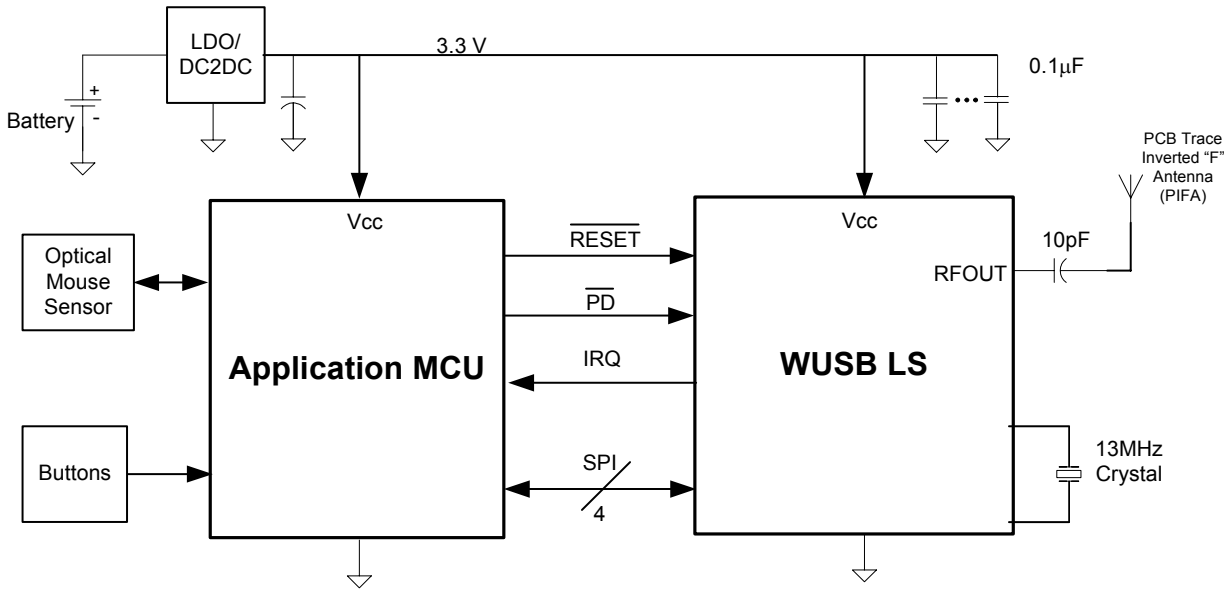


Figure 6-1. CYWUSB6932 Transmit-Only Battery-Powered Device

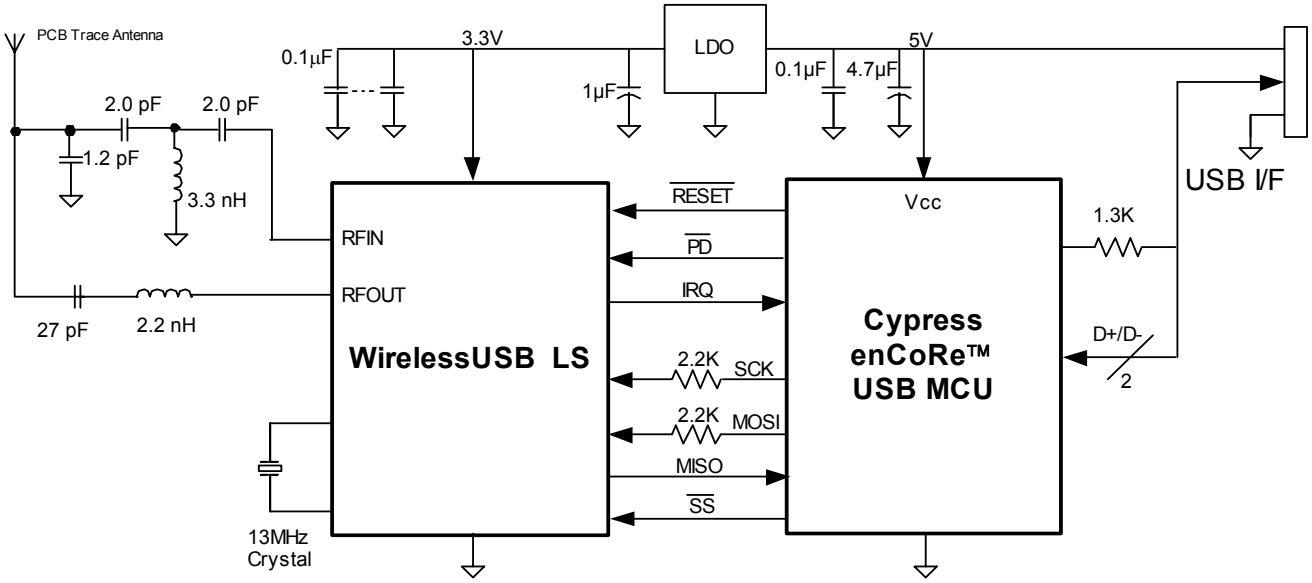


Figure 6-2. CYWUSB6934 USB Bridge Transceiver

7.0 Register Descriptions

Table 7-1 displays the list of registers inside the CYWUSB6932/CYWUSB6934 ICs that are addressable through the SPI interface. All registers are read and writable, except where noted.

Table 7-1. CYWUSB6932/CYWUSB6934 Register Map[2]

Register Name	Mnemonic	Address	Page	Default	Access
Revision ID	REG_ID	0x00	8	0x07	RO
Control	REG_CONTROL	0x03	8	0x00	RW
Data Rate	REG_DATA_RATE	0x04	9	0x00	RW
Configuration	REG_CONFIG	0x05	9	0x01	RW
SERDES Control	REG_SERDES_CTL	0x06	10	0x03	RW
Receive SERDES Interrupt Enable	REG_RX_INT_EN	0x07[1]	11	0x00	RW
Receive SERDES Interrupt Status	REG_RX_INT_STAT	0x08[1]	12	0x00	RO
Receive SERDES Data A	REG_RX_DATA_A	0x09[1]	13	0x00	RO
Receive SERDES Valid A	REG_RX_VALID_A	0x0A[1]	13	0x00	RO
Receive SERDES Data B	REG_RX_DATA_B	0x0B[1]	13	0x00	RO
Receive SERDES Valid B	REG_RX_VALID_B	0x0C[1]	13	0x00	RO
Transmit SERDES Interrupt Enable	REG_TX_INT_EN	0x0D	14	0x00	RW
Transmit SERDES Interrupt Status	REG_TX_INT_STAT	0x0E	15	0x00	RO
Transmit SERDES Data	REG_TX_DATA	0x0F	16	0x00	RW
Transmit SERDES Valid	REG_TX_VALID	0x10	16	0x00	RW
PN Code	REG_PN_CODE	0x18–0x11	16	0x1E8B6A3DE0E9B222	RW
Threshold Low	REG_THRESHOLD_L	0x19[1]	17	0x08	RW
Threshold High	REG_THRESHOLD_H	0x1A[1]	17	0x38	RW
Wake Enable	REG_WAKE_EN	0x1C	17	0x00	RW
Wake Status	REG_WAKE_STAT	0x1D	18	0x01	RO
Analog Control	REG_ANALOG_CTL	0x20	18	0x04	RW
Channel	REG_CHANNEL	0x21	19	0x00	RW
Receive Signal Strength Indicator	REG_RSSI	0x22[1]	19	0x00	RO
PA Bias	REG_PA	0x23	19	0x00	RW
Crystal Adjust	REG_CRYSTAL_ADJ	0x24	20	0x00	RW
VCO Calibration	REG_VCO_CAL	0x26	20	0x00	RW
Reg Power Control	REG_PWR_CTL	0x2E	20	0x00	RW
Carrier Detect	REG_CARRIER_DETECT	0x2F	21	0x00	RW
Clock Manual	REG_CLOCK_MANUAL	0x32	21	0x00	RW
Clock Enable	REG_CLOCK_ENABLE	0x33	21	0x00	RW
Synthesizer Lock Count	REG_SYN_LOCK_CNT	0x38	21	0x64	RW
Manufacturing ID	REG_MID	0x3C–0x3F	21	–	RO

Notes:
1. Register not applicable to CYWUSB6932.
2. All registers are accessed Little Endian.

Figure 7-1. Revision ID Register

Addr: 0x00			REG_ID				Default: 0x07
7	6	5	4	3	2	1	0
Silicon ID				Product ID			

Bit	Name	Description
7:4	Silicon ID	These are the Silicon ID revision bits. 0000 = Rev A, 0001 = Rev B, etc. These bits are read-only.
3:0	Product ID	These are the Product ID revision bits. Fixed at value 0111. These bits are read-only.

Figure 7-2. Control

Addr: 0x03			REG_CONTROL				Default: 0x00
7	6	5	4	3	2	1	0
RX Enable	TX Enable	PN Code Select	Bypass Internal Syn Lock Signal	Auto Internal PA Disable	Internal PA Enable	Reserved	Reserved

Bit	Name	Description
7	RX Enable	The Receive Enable bit is used to place the IC in receive mode. 1 = Receive Enabled 0 = Receive Disabled
6	TX Enable	The Transmit Enable bit is used to place the IC in transmit mode. 1 = Transmit Enabled 0 = Transmit Disabled
5	PN Code Select	The Pseudo-Noise Code Select bit selects between the upper or lower half of the 64 chips/bit PN code. 1 = 32 Most Significant Bits of PN code are used 0 = 32 Least Significant Bits of PN code are used This bit applies only when the Code Width bit is set to 32 chips/bit PN codes (Reg 0x04, bit 2=1).
4	Bypass Internal Syn Lock Signal	This bit controls whether the state machine waits for the internal Syn Lock Signal before waiting for the amount of time specified in the Syn Lock Count register (Reg 0x38), in units of 2 μs. If the internal Syn Lock Signal is used then set Syn Lock Count to 25 to provide additional assurance that the synthesizer has settled. 1 = Bypass the Internal Syn Lock Signal and wait the amount of time in Syn Lock Count register (Reg 0x38) 0 = Wait for the Syn Lock Signal and then wait the amount of time specified in Syn Lock Count register (Reg 0x38) It is recommended that the application MCU sets this bit to 1 in order to guarantee a consistent settle time for the synthesizer.
3	Auto Internal PA Disable	The Auto Internal PA Disable bit is used to determine the method of controlling the Internal Power Amplifier. The two options are automatic control by the baseband or by firmware through register writes. For external PA usage, please see the description of the REG_ANALOG_CTL register (Reg 0x20). 1 = Register controlled Internal PA Enable 0 = Auto controlled Internal PA Enable When this bit is set to 1, the enabled state of the Internal PA is directly controlled by bit Internal PA Enable (Reg 0x03, bit 2). It is recommended that this bit is set to 0, leaving the PA control to the baseband.
2	Internal PA Enable	The Internal PA Enable bit is used to enable or disable the Internal Power Amplifier. 1 = Internal Power Amplifier Enabled 0 = Internal Power Amplifier Disabled This bit only applies when the Auto Internal PA Disable bit is selected (Reg 0x03, bit 3=1), otherwise this bit is don't care.
1	Reserved	This bit is reserved and should be written with a zero.
0	Reserved	This bit is reserved and should be written with a zero.

Addr: 0x04			REG_DATA_RATE				Default: 0x00
7	6	5	4	3	2	1	0
Reserved					Code Width	Data Rate	Sample Rate

Figure 7-3. Data Rate

Bit	Name	Description
7:3	Reserved	These bits are reserved and should be written with zeroes.
2[3]	Code Width	The Code Width bit is used to select between 32 chips/bit and 64 chips/bit PN codes. 1 = 32 chips/bit PN codes 0 = 64 chips/bit PN codes The number of chips/bit used impacts a number of factors such as data throughput, range and robustness to interference. By choosing a 32 chips/bit PN-code, the data throughput can be doubled or even quadrupled (when double data rate is set). A 64 chips/bit PN code offers improved range over its 32 chips/bit counterpart as well as more robustness to interference. By selecting to use a 32 chips/bit PN code a number of other register bits are impacted and need to be addressed. These are PN Code Select (Reg 0x03, bit 5), Data Rate (Reg 0x04, bit 1), and Sample Rate (Reg 0x04, bit 0).
1[3]	Data Rate	The Data Rate bit allows the user to select Double Data Rate mode of operation which delivers a raw data rate of 62.5kbits/sec. 1 = Double Data Rate - 2 bits per PN code (No odd bit transmissions) 0 = Normal Data Rate - 1 bit per PN code This bit is applicable only when using 32 chips/bit PN codes which can be selected by setting the Code Width bit (Reg 0x04, bit 2=1). When using Double Data Rate, the raw data throughput is 62.5 kbits/sec because every 32 chips/bit PN code is interpreted as 2 bits of data. When using this mode a single 64 chips/bit PN code is placed in the PN code register. This 64 chips/bit PN code is then split into two and used by the baseband to offer the Double Data Rate capability. When using Normal Data Rate, the raw data throughput is 32kbits/sec. Additionally, Normal Data Rate enables the user to potentially correlate data using two differing 32 chips/bit PN codes.
0[3]	Sample Rate	The Sample Rate bit allows the use of the 12x sampling when using 32 chips/bit PN codes and Normal Data Rate. 1 = 12x Oversampling 0 = 6x Oversampling Using 12x oversampling improves the correlators receive sensitivity. When using 64 chips/bit PN codes or Double Data Rate this bit is don't care. The only time when 12x oversampling can be selected is when a 32 chips/bit PN code is being used with Normal Data Rate.

Figure 7-4. Configuration

Addr: 0x05			REG_CONFIG				Default: 0x01
7	6	5	4	3	2	1	0
Reserved						IRQ Pin Select	

Bit	Name	Description
7:2	Reserved	These bits are reserved and should be written with zeroes.
1:0	IRQ Pin Select	The Interrupt Request Pin Select bits are used to determine the drive method of the IRQ pin. 11 = Open Source (IRQ asserted = 1, IRQ deasserted = Hi-Z) 10 = Open Drain (IRQ asserted = 0, IRQ deasserted = Hi-Z) 01 = CMOS (IRQ asserted = 1, IRQ deasserted = 0) 00 = CMOS Inverted (IRQ asserted = 0, IRQ deasserted = 1)

Note:
3. The following Reg 0x04, bits 2:0 values are not valid:
- 001 – Not Valid
- 010 – Not Valid
- 011 – Not Valid
- 111 – Not Valid.

Addr: 0x06			REG_SERDES_CTL				Default: 0x03
7	6	5	4	3	2	1	0
Reserved				SERDES Enable	EOF Length		

Figure 7-5. SERDES Control

Bit	Name	Description
7:4	Reserved	These bits are reserved and should be written with zeroes.
3	SERDES Enable	The SERDES Enable bit is used to switch between bit-serial mode and SERDES mode. 1 = SERDES enabled. 0 = SERDES disabled, bit-serial mode enabled. When the SERDES is enabled data can be written to and read from the IC one byte at a time, through the use of the SERDES Data registers. The bit-serial mode requires bits to be written one bit at a time through the use of the DIO/DIOVAL pins, refer to section 3.2. It is recommended that SERDES mode be used to avoid the need to manage the timing required by the bit-serial mode.
2:0	EOF Length	The End of Frame Length bits are used to set the number of sequential bit times for an inter-frame gap without valid data before an EOF event will be generated. When in receive mode and a valid bit has been received the EOF event can then be identified by the number of bit times that expire without correlating any new data. The EOF event causes data to be moved to the proper SERDES Data Register and can also be used to generate interrupts. If 0 is the EOF length, an EOF condition will occur at the first invalid bit after a valid reception.

Addr: 0x07			REG_RX_INT_EN				Default: 0x00
7	6	5	4	3	2	1	0
Underflow B	Overflow B	EOF B	Full B	Underflow A	Overflow A	EOF A	Full A

Figure 7-6. Receive SERDES Interrupt Enable

Bit	Name	Description
7	Underflow B	The Underflow B bit is used to enable the interrupt associated with an underflow condition with the Receive SERDES Data B register (Reg 0x0B) 1 = Underflow B interrupt enabled for Receive SERDES Data B 0 = Underflow B interrupt disabled for Receive SERDES Data B An underflow condition occurs when attempting to read the Receive SERDES Data B register (Reg 0x0B) when it is empty.
6	Overflow B	The Overflow B bit is used to enable the interrupt associated with an overflow condition with the Receive SERDES Data B register (Reg 0x0B) 1 = Overflow B interrupt enabled for Receive SERDES Data B 0 = Overflow B interrupt disabled for Receive SERDES Data B An overflow condition occurs when new received data is written into the Receive SERDES Data B register (Reg 0x0B) before the prior data is read out.
5	EOF B	The End of Frame B bit is used to enable the interrupt associated with the Channel B Receiver EOF condition. 1 = EOF B interrupt enabled for Channel B Receiver. 0 = EOF B interrupt disabled for Channel B Receiver. The EOF IRQ asserts during an End of Frame condition. End of Frame conditions occur after at least one bit has been detected, and then the number of invalid bits in the frame exceeds the number in the EOF length field. If 0 is the EOF length, and EOF condition will occur at the first invalid bit after a valid reception. This IRQ is cleared by reading the receive status register
4	Full B	The Full B bit is used to enable the interrupt associated with the Receive SERDES Data B register (Reg 0x0B) having data placed in it. 1 = Full B interrupt enabled for Receive SERDES Data B 0 = Full B interrupt disabled for Receive SERDES Data B A Full B condition occurs when data is transferred from the Channel B Receiver into the Receive SERDES Data B register (Reg 0x0B). This could occur when a complete byte is received or when an EOF event occurs whether or not a complete byte has been received.
3	Underflow A	The Underflow A bit is used to enable the interrupt associated with an underflow condition with the Receive SERDES Data A register (Reg 0x09) 1 = Underflow A interrupt enabled for Receive SERDES Data A 0 = Underflow A interrupt disabled for Receive SERDES Data A An underflow condition occurs when attempting to read the Receive SERDES Data A register (Reg 0x09) when it is empty.
2	Overflow A	The Overflow A bit is used to enable the interrupt associated with an overflow condition with the Receive SERDES Data A register (0x09) 1 = Overflow A interrupt enabled for Receive SERDES Data A 0 = Overflow A interrupt disabled for Receive SERDES Data A An overflow condition occurs when new receive data is written into the Receive SERDES Data A register (Reg 0x09) before the prior data is read out.
1	EOF A	The End of Frame A bit is used to enable the interrupt associated with an End of Frame condition with the Channel A Receiver. 1 = EOF A interrupt enabled for Channel A Receiver. 0 = EOF A interrupt disabled for Channel A Receiver. The EOF IRQ asserts during an End of Frame condition. End of Frame conditions occur after at least one bit has been detected, and then the number of invalid bits in a frame exceeds the number in the EOF length field. If 0 is the EOF length, an EOF condition will occur at the first invalid bit after a valid reception. This IRQ is cleared by reading the receive status register.
0	Full A	The Full A bit is used to enable the interrupt associated with the Receive SERDES Data A register (0x09) having data written into it. 1 = Full A interrupt enabled for Receive SERDES Data A 0 = Full A interrupt disabled for Receive SERDES Data A A Full A condition occurs when data is transferred from the Channel A Receiver into the Receive SERDES Data A register (Reg 0x09). This could occur when a complete byte is received or when an EOF event occurs whether or not a complete byte has been received.

Addr: 0x08			REG_RX_INT_STAT				Default: 0x00
7	6	5	4	3	2	1	0
Valid B	Flow Violation B	EOF B	Full B	Valid A	Flow Violation A	EOF A	Full A

Figure 7-7. Receive SERDES Interrupt Status[4]

Bit	Name	Description
7	Valid B	The Valid B bit is true when all the bits in the Receive SERDES Data B register (Reg 0x0B) are valid. 1 = All bits are valid for Receive SERDES Data B. 0 = Not all bits are valid for Receive SERDES Data B. When data is written into the Receive SERDES Data B register (Reg 0x0B) this bit is set if all of the bits within the byte that has been written are valid. This bit cannot generate an interrupt.
6	Flow Violation B	The Flow Violation B bit is used to signal whether an overflow or underflow condition has occurred for the Receive SERDES Data B register (Reg 0x0B). 1 = Overflow/underflow interrupt pending for Receive SERDES Data B. 0 = No overflow/underflow interrupt pending for Receive SERDES Data B. Overflow conditions occur when the radio loads new data into the Receive SERDES Data B register (Reg 0x0B) before the prior data has been read. Underflow conditions occur when trying to read the Receive SERDES Data B register (Reg 0x0B) when the register is empty. This bit is cleared by reading the Receive Interrupt Status register (Reg 0x08)
5	EOF B	The End of Frame B bit is used to signal whether an EOF event has occurred on the Channel B receive. 1 = EOF interrupt pending for Channel B. 0 = No EOF interrupt pending for Channel B. An EOF condition occurs for the Channel B Receiver when receive has begun and then the number of bit times specified in the SERDES Control register (Reg 0x06) elapse without any valid bits being received. This bit is cleared by reading the Receive Interrupt Status register (Reg 0x08)
4	Full B	The Full B bit is used to signal when the Receive SERDES Data B register (Reg 0x0B) is filled with data. 1 = Receive SERDES Data B full interrupt pending. 0 = No Receive SERDES Data B full interrupt pending. A Full B condition occurs when data is transferred from the Channel B Receiver into the Receive SERDES Data B register (Reg 0x0B). This could occur when a complete byte is received or when an EOF event occurs whether or not a complete byte has been received.
3	Valid A	The Valid A bit is true when all of the bits in the Receive SERDES Data A Register (Reg 0x09) are valid. 1 = All bits are valid for Receive SERDES Data A. 0 = Not all bits are valid for Receive SERDES Data A. When data is written into the Receive SERDES Data A register (Reg 0x09) this bit is set if all of the bits within the byte that has been written are valid. This bit cannot generate an interrupt.
2	Flow Violation A	The Flow Violation A bit is used to signal whether an overflow or underflow condition has occurred for the Receive SERDES Data A register (Reg 0x09). 1 = Overflow/underflow interrupt pending for Receive SERDES Data A. 0 = No overflow/underflow interrupt pending for Receive SERDES Data A. Overflow conditions occur when the radio loads new data into the Receive SERDES Data A register (Reg 0x09) before the prior data has been read. Underflow conditions occur when trying to read the Receive SERDES Data A register (Reg 0x09) when the register is empty. This bit is cleared by reading the Receive Interrupt Status register (Reg 0x08)
1	EOF A	The End of Frame A bit is used to signal whether an EOF event has occurred on the Channel A receive. 1 = EOF interrupt pending for Channel A. 0 = No EOF interrupt pending for Channel A. An EOF condition occurs for the Channel A Receiver when receive has begun and then the number of bit times specified in the SERDES Control register (0x06) elapse without any valid bits being received. This bit is cleared by reading the Receive Interrupt Status register (Reg 0x08).
0	Full A	The Full A bit is used to signal when the Receive SERDES Data A register (Reg 0x09) is filled with data. 1 = Receive SERDES Data A full interrupt pending. 0 = No Receive SERDES Data A full interrupt pending. A Full A condition occurs when data is transferred from the Channel A Receiver into the Receive SERDES Data A Register (Reg 0x09). This could occur when a complete byte is received or when an EOF event occurs whether or not a complete byte has been received.

Note:

4. All status bits are set and readable in the registers regardless of IRQ enable status. This allows a polling scheme to be implemented without enabling IRQs. The status bits are affected by TX Enable and RX Enable (Reg 0x03, bits 7:6). For example, the receive status will read 0 if the IC is not in receive mode. These registers are read-only.

Addr: 0x09			REG_RX_DATA_A				Default: 0x00
7	6	5	4	3	2	1	0
Data							

Figure 7-8. Receive SERDES Data A

Bit	Name	Description
7:0	Data	Received Data for Channel A. The over-the-air received order is bit 0 followed by bit 1, followed by bit 2, followed by bit 3, followed by bit 4, followed by bit 5, followed by bit 6, followed by bit 7. This register is read-only.

Addr: 0x0A			REG_RX_VALID_A				Default: 0x00
7	6	5	4	3	2	1	0
Valid							

Figure 7-9. Receive SERDES Valid A

Bit	Name	Description
7:0	Valid	These bits indicate which of the bits in the Receive SERDES Data A register (Reg 0x09) are valid. A “1” indicates that the corresponding data bit is valid for Channel A. If the Valid Data bit is set in the Receive Interrupt Status register (Reg 0x08) all eight bits in the Receive SERDES Data A register (Reg 0x0A) are valid. Therefore, it is not necessary to read the Receive SERDES Valid A register (Reg 0x0C). This register is read-only.

Addr: 0x0B			REG_RX_DATA_B				Default: 0x00
7	6	5	4	3	2	1	0
Data							

Figure 7-10. Receive SERDES Data B

Bit	Name	Description
7:0	Data	Received Data for Channel B. The over-the-air received order is bit 0 followed by bit 1, followed by bit 2, followed by bit 3, followed by bit 4, followed by bit 5, followed by bit 6, followed by bit 7. This register is read-only.

Figure 7-11. Receive SERDES Valid B

Addr: 0x0C			REG_RX_VALID_B				Default: 0x00
7	6	5	4	3	2	1	0
Valid							

Bit	Name	Description
7:0	Valid	These bits indicate which of the bits in the Receive SERDES Data B register (Reg 0x0B) are valid. A “1” indicates that the corresponding data bit is valid for Channel B. If the Valid Data bit is set in the Receive Interrupt Status register (0x08) all eight bits in the Receive SERDES Data B register (Reg 0x0B) are valid. Therefore, it is not necessary to read the Receive SERDES Valid B register (Reg 0x0C). This register is read-only.

Addr: 0x0D			REG_TX_INT_EN				Default: 0x00
7	6	5	4	3	2	1	0
Reserved				Underflow	Overflow	Done	Empty

Figure 7-12. Transmit SERDES Interrupt Enable

Bit	Name	Description
7:4	Reserved	These bits are reserved and should be written with zeroes.
3	Underflow	The Underflow bit is used to enable the interrupt associated with an underflow condition associated with the Transmit SERDES Data register (Reg 0x0F) 1 = Underflow interrupt enabled. 0 = Underflow interrupt disabled. An underflow condition occurs when attempting to transmit while the Transmit SERDES Data register (Reg 0x0F) does not have any data.
2	Overflow	The Overflow bit is used to enabled the interrupt associated with an overflow condition with the Transmit SERDES Data register (0x0F). 1 = Overflow interrupt enabled. 0 = Overflow interrupt disabled. An overflow condition occurs when attempting to write new data to the Transmit SERDES Data register (Reg 0x0F) before the preceding data has been transferred to the transmit shift register.
1	Done	The Done bit is used to enable the interrupt that signals the end of the transmission of data. 1 = Done interrupt enabled. 0 = Done interrupt disabled. The Done condition occurs when the Transmit SERDES Data register (Reg 0x0F) has transmitted all of its data and there is no more data for it to transmit.
0	Empty	The Empty bit is used to enable the interrupt that signals when the Transmit SERDES register (Reg 0x0F) is empty. 1 = Empty interrupt enabled. 0 = Empty interrupt disabled. The Empty condition occurs when the Transmit SERDES Data register (Reg 0x0F) is loaded into the transmit buffer and it's safe to load the next byte

Addr: 0x0E			REG_TX_INT_STAT			Default: 0x00	
7	6	5	4	3	2	1	0
Reserved				Underflow	Overflow	Done	Empty

Figure 7-13. Transmit SERDES Interrupt Status[5]

Bit	Name	Description
7:4	Reserved	These bits are reserved. This register is read-only.
3	Underflow	The Underflow bit is used to signal when an underflow condition associated with the Transmit SERDES Data register (Reg 0x0F) has occurred. 1 = Underflow Interrupt pending. 0 = No Underflow Interrupt pending. This IRQ will assert during an underflow condition to the Transmit SERDES Data register (Reg 0x0F). An underflow occurs when the transmitter is ready to sample transmit data, but there is no data ready in the Transmit SERDES Data register (Reg 0x0F). This will only assert after the transmitter has transmitted at least one bit. This bit is cleared by reading the Transmit Interrupt Status register (Reg 0x0E).
2	Overflow	The Overflow bit is used to signal when an overflow condition associated with the Transmit SERDES Data register (0x0F) has occurred. 1 = Overflow Interrupt pending. 0 = No Overflow Interrupt pending. This IRQ will assert during an overflow condition to the Transmit SERDES Data register (Reg 0x0F). An overflow occurs when the new data is loaded into the Transmit SERDES Data register (Reg 0x0F) before the previous data has been sent. This bit is cleared by reading the Transmit Interrupt Status register (Reg 0x0E).
1	Done	The Done bit is used to signal the end of a data transmission. 1 = Done Interrupt pending. 0 = No Done Interrupt pending. This IRQ will assert when the data is finished sending a byte of data and there is no more data to be sent. This will only assert after the transmitter has transmitted as least one bit. This bit is cleared by reading the Transmit Interrupt Status register (Reg 0x0E)
0	Empty	The Empty bit is used to signal when the Transmit SERDES Data register (Reg 0x0F) has been emptied. 1 = Empty Interrupt pending. 0 = No Empty Interrupt pending. This IRQ will assert when the transmit serdes is empty. When this IRQ is asserted it is ok to write to the Transmit SERDES Data register (Reg 0x0F). Writing the Transmit SERDES Data register (Reg 0x0F) will clear this IRQ. It will be set when the data is loaded into the transmitter, and it is ok to write new data.

Note:

5. All status bits are set and readable in the registers regardless of IRQ enable status. This allows a polling scheme to be implemented without enabling IRQs. The status bits are affected by the TX Enable and RX Enable (Reg 0x03, bits 7:6). For example, the transmit status will read 0 if the IC is not in transmit mode. These registers are read-only.

Addr: 0x0F			REG_TX_DATA				Default: 0x00
7	6	5	4	3	2	1	0
Data							

Figure 7-14. Transmit SERDES Data

Bit	Name	Description
7:0	Data	Transmit Data. The over-the-air transmitted order is bit 0 followed by bit 1, followed by bit 2, followed by bit 3, followed by bit 4, followed by bit 5, followed by bit 6, followed by bit 7.

Addr: 0x10			REG_TX_VALID				Default: 0x00
7	6	5	4	3	2	1	0
Valid							

Figure 7-15. Transmit SERDES Valid

Bit	Name	Description
7:0	Valid[6]	The Valid bits are used to determine which of the bits in the Transmit SERDES Data register (reg 0x0F) are valid. 1 = Valid transmit bit. 0 = Invalid transmit bit.

Addr: 0x11-18								REG_PN_CODE																Default: 0x1E8B6A3DE0E9B222							
63	62	61	60	59	58	57	56	55	54	53	52	51	50	49	48	47	46	45	44	43	42	41	40	39	38	37	36	35	34	33	32
Address 0x18								Address 0x17								Address 0x16								Address 0x15							

Figure 7-16. PN Code

31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Address 0x14								Address 0x13								Address 0x12								Address 0x11							

Bit	Name	Description
63:0	PN Codes	The value inside the 8 byte PN code register is used as the spreading code for DSSS communication. All 8 bytes can be used together for 64 chips/bit PN code communication, or the registers can be split into two sets of 32 chips/bit PN codes and these can be used alone or with each other to accomplish faster data rates. Not any 64 chips/bit value can be used as a PN code as there are certain characteristics that are needed to minimize the possibility of multiple PN codes interfering with each other or the possibility of invalid correlation. The over-the-air order is bit 0 followed by bit 1... followed by bit 62, followed by bit 63.

Note:

6. The Valid bit in the Transmit SERDES Valid register (Reg 0x10) is used to mark whether the radio will send data or preamble during that bit time of the data byte. Data is sent LSB first. The SERDES will continue to send data until there are no more VALID bits in the shifter. For example, writing 0x0F to the Transmit SERDES Valid register (Reg 0x10) will send half a byte.

Addr: 0x19			REG_THRESHOLD_L				Default: 0x08
7	6	5	4	3	2	1	0
Reserved	Threshold Low						

Figure 7-17. Threshold Low

Bit	Name	Description
7	Reserved	This bit is reserved and should be written with zero.
6:0	Threshold Low	The Threshold Low value is used to determine the number of missed chips allowed when attempting to correlate a single data bit of value ‘0’. A perfect reception of a data bit of ‘0’ with a 64 chips/bit PN code would result in zero correlation matches, meaning the exact inverse of the PN code has been received. By setting the Threshold Low value to 0x08 for example, up to eight chips can be erroneous while still identifying the value of the received data bit. This value along with the Threshold High value determine the correlator count values for logic ‘1’ and logic ‘0’. The threshold values used determine the sensitivity of the receiver to interference and the dependability of the received data. By allowing a minimal number of erroneous chips the dependability of the received data increases while the robustness to interference decreases. On the other hand increasing the maximum number of missed chips means reduced data integrity but increased robustness to interference and increased range.

Addr: 0x1A			REG_THRESHOLD_H				Default: 0x38
7	6	5	4	3	2	1	0
Reserved	Threshold High						

Figure 7-18. Threshold High

Bit	Name	Description
7	Reserved	This bit is reserved and should be written with zero.
6:0	Threshold High	The Threshold High value is used to determine the number of matched chips allowed when attempting to correlate a single data bit of value ‘1’. A perfect reception of a data bit of ‘1’ with a 64 chips/bit or a 32 chips/bit PN code would result in 64 chips/bit or 32 chips/bit correlation matches, respectively, meaning every bit was received perfectly. By setting the Threshold High value to 0x38 (64-8) for example, up to eight chips can be erroneous while still identifying the value of the received data bit. This value along with the Threshold Low value determine the correlator count values for logic ‘1’ and logic ‘0’. The threshold values used determine the sensitivity of the receiver to interference and the dependability of the received data. By allowing a minimal number of erroneous chips the dependability of the received data increases while the robustness to interference decreases. On the other hand increasing the maximum number of missed chips means reduced data integrity but increased robustness to interference and increased range.

Addr: 0x1C			REG_WAKE_EN				Default: 0x00
7	6	5	4	3	2	1	0
Reserved							Wakeup Enable

Figure 7-19. Wake Enable

Bit	Name	Description
7:1	Reserved	These bits are reserved and should be written with zeroes.
0	Wakeup Enable	Wakeup interrupt enable. 0 = disabled 1 = enabled A wakeup event is triggered when the $\overline{PD}$ pin is deasserted and once the IC is ready to receive SPI communications.

Addr: 0x1D			REG_WAKE_STAT				Default: 0x01
7	6	5	4	3	2	1	0
Reserved							Wakeup Status

Figure 7-20. Wake Status

Bit	Name	Description
7:1	Reserved	These bits are reserved. This register is read-only.
0	Wakeup Status	Wakeup status. 0 = Wake interrupt not pending 1 = Wake interrupt pending This IRQ will assert when a wakeup condition occurs. This bit is cleared by reading the Wake Status register (Reg 0x1D). This register is read-only.

Addr: 0x20			REG_ANALOG_CTL				Default: 0x00
7	6	5	4	3	2	1	0
Reserved	Reg Write Control	MID Read Enable	Reserved	Reserved	PA Output Enable	PA Invert	Reset

Figure 7-21. Analog Control

Bit	Name	Description
7	Reserved	This bit is reserved and should be written with zero.
6	Reg Write Control	Enables write access to Reg 0x2E and Reg 0x2F. 1 = Enables write access to Reg 0x2E and Reg 0x2F 0 = Reg 0x2E and Reg 0x2F are read-only
5	MID Read Enable	The MID Read Enable bit must be set to read the contents of the Manufacturing ID register (Reg 0x3C-0x3F). Enabling the Manufacturing ID register (Reg 0x3C-0x3F) consumes power. This bit should only be set when reading the contents of the Manufacturing ID register (Reg 0x3C-0x3F). 1 = Enables read of MID registers 0 = Disables read of MID registers
4:3	Reserved	These bits are reserved and should be written with zeroes.
2	PA Output Enable	The Power Amplifier Output Enable bit is used to enable the PACTL pin for control of an external power amplifier. 1 = PA Control Output Enabled on PACTL pin 0 = PA Control Output Disabled on PACTL pin
1	PA Invert	The Power Amplifier Invert bit is used to specify the polarity of the PACTL signal when the PA Output Enable bit is set high. PA Output Enable and PA Invert cannot be simultaneously changed. 1 = PACTL active low 0 = PACTL active high
0	Reset	The Reset bit is used to generate a self-clearing device reset. 1 = Device Reset. All registers are restored to their default values. 0 = No Device Reset.

Addr: 0x21			REG_CHANNEL				Default: 0x00
7	6	5	4	3	2	1	0
Reserved	Channel						

Figure 7-22. Channel

Bit	Name	Description
7	Reserved	This bit is reserved and should be written with zero.
6:0	Channel	The Channel register (Reg 0x21) is used to determine the Synthesizer frequency. A value of 2 corresponds to a communication frequency of 2.402 GHz, while a value of 79 corresponds to a frequency of 2.479GHz. The channels are separated from each other by 1 MHz intervals. Limit application usage to channels 2-79 to adhere to FCC regulations. FCC regulations require that channels 0 and 1 and any channel greater than 79 be avoided. Use of other channels may be restricted by other regulatory agencies. The application MCU must ensure that this register is modified before transmitting data over the air for the first time.

Addr: 0x22			REG_RSSI				Default: 0x00
7	6	5	4	3	2	1	0
Reserved		Valid	RSSI				

Figure 7-23. Receive Signal Strength Indicator (RSSI)[7]

Bit	Name	Description
7:6	Reserved	These bits are reserved. This register is read-only.
5	Valid	The Valid bit indicates whether the RSSI value in bits [4:0] are valid. This register is Read Only. 1 = RSSI value is valid 0 = RSSI value is invalid
4:0	RSSI	The Receive Strength Signal Indicator (RSSI) value indicates the strength of the received signal. This is a read only value with the higher values indicating stronger received signals meaning more reliable transmissions.

Addr: 0x23			REG_PA				Default: 0x00
7	6	5	4	3	2	1	0
Reserved					PA Bias		

Figure 7-24. PA Bias

Bit	Name	Description
7:3	Reserved	These bits are reserved and should be written with zeroes.
2:0	PA Bias	The Power Amplifier Bias (PA Bias) bits are used to set the transmit power of the IC through increasing (values up to 7) or decreasing (values down to 0) the gain of the on-chip Power Amplifier. The higher the register value the higher the transmit power. By changing the PA Bias value signal strength management functions can be accomplished. For general purpose communication a value of 7 is recommended.

Note:
7. The RSSI will collect a single value each time the part is put into receive mode via Control register (Reg 0x03, bit 7=1). See *Section 4.7* for more details.

Addr: 0x24		REG_CRYSTAL_ADJ					Default: 0x00
7	6	5	4	3	2	1	0
Reserved	Clock Output Disable	Crystal Adjust					

Figure 7-25. Crystal Adjust

Bit	Name	Description
7	Reserved	This bit is reserved and should be written with zero.
6	Clock Output Disable	The Clock Output Disable bit disables the 13 MHz clock driven on the X13OUT pin. 1 = No 13 MHz clock driven externally. 0 = 13 MHz clock driven externally. If the 13 MHz clock is driven on the X13OUT pin then receive sensitivity will be reduced by –4 dBm on channels 5+13*n*. By default the 13 MHz clock output pin is enabled. This pin is useful for adjusting the 13 MHz clock, but it interfere with every 13th channel beginning with 2.405GHz channel. Therefore, it is recommended that the 13 MHz clock output pin be disabled when not in use.
5:0	Crystal Adjust	The Crystal Adjust value is used to calibrate the on-chip parallel load capacitance supplied to the crystal. Each increment of the Crystal Adjust value typically adds 0.135 pF of parallel load capacitance. The total range is 8.5 pF, starting at 8.65 pF. These numbers do not include PCB parasitics, which can add an additional 1–2 pF.

Addr: 0x26		REG_VCO_CAL					Default: 0x00
7	6	5	4	3	2	1	0
VCO Slope Enable		Reserved					

Figure 7-26. VCO Calibration

Bit	Name	Description
7:6	VCO Slope Enable (Write-Only)	The Voltage Controlled Oscillator (VCO) Slope Enable bits are used to specify the amount of variance automatically added to the VCO. 11 = –5/+5 VCO adjust. The application MCU must configure this option during initialization. 10 = –2/+3 VCO adjust. 01 = Reserved. 00 = No VCO adjust. These bits are undefined for read operations.
5:0	Reserved	These bits are reserved and should be written with zeroes.

Addr: 0x2E		REG_PWR_CTL					Default: 0x00
7	6	5	4	3	2	1	0
Reg Power Control	Reserved						

Figure 7-27. Reg Power Control

Bit	Name	Description
7	Reg Power Control	When set, this bit disables unused circuitry and saves radio power. The user must set Reg 0x20, bit 6=1 to enable writes to Reg 0x2E. The application MCU must set this bit during initialization.
6:0	Reserved	These bits are reserved and should be written with zeroes.

Addr: 0x2F			REG_CARRIER_DETECT				Default: 0x00
7	6	5	4	3	2	1	0
Carrier Detect Override	Reserved						

Figure 7-28. Carrier Detect

Bit	Name	Description
7	Carrier Detect Override	When set, this bit overrides carrier detect. The user must set Reg 0x20, bit 6=1 to enable writes to Reg 0x2F.
6:0	Reserved	These bits are reserved and should be written with zeroes.

Addr: 0x32			REG_CLOCK_MANUAL				Default: 0x00
7	6	5	4	3	2	1	0
Manual Clock Overrides							

Figure 7-29. Clock Manual

Bit	Name	Description
7:0	Manual Clock Overrides	This register must be written with 0x41 after reset for correct operation

Addr: 0x33			REG_CLOCK_ENABLE				Default: 0x00
7	6	5	4	3	2	1	0
Manual Clock Enables							

Figure 7-30. Clock Enable

Bit	Name	Description
7:0	Manual Clock Enables	This register must be written with 0x41 after reset for correct operation

Addr: 0x38			REG_SYN_LOCK_CNT				Default: 0x64
7	6	5	4	3	2	1	0
Count							

Figure 7-31. Synthesizer Lock Count

Bit	Name	Description
7:0	Count	Determines the length of delay in 2µs increments for the synthesizer to lock when auto synthesizer is enabled via Control register (0x03, bit 1=0) and not using the PLL lock signal. The default register setting is typically sufficient.

Addr: 0x3C-3F								REG_MID																							
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Address 0x3F								Address 0x3E								Address 0x3D								Address 0x3C							

Figure 7-32. Manufacturing ID

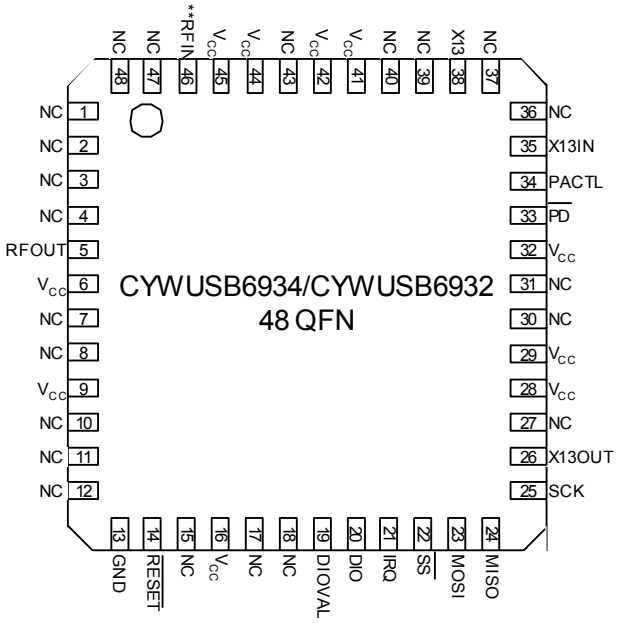
Bit	Name	Description
31:30	Address[31:30]	These bits are read back as zeroes.
29:0	Address[29:0]	These bits are the Manufacturing ID (MID) for each IC. The contents of these bits cannot be read unless the MID Read Enable bit (bit 5) is set in the Analog Control register (Reg 0x20). Enabling the Manufacturing ID register (Reg 0x3C-0x3F) consumes power. The MID Read Enable bit in the Analog Control register (Reg 0x20, bit 5) should only be set when reading the contents of the Manufacturing ID register (Reg 0x3C-0x3F). This register is read-only.

8.0 Pin Definitions

Table 8-1. Pin Description Table for the CYWUSB6932/CYWUSB6934

Pin QFN	Name	Type	Default	Description
46	RFIN	Input	Input	**RF Input**. Modulated RF signal received (CYWUSB6934 only).
5	RFOUT	Output	N/A	**RF Output**. Modulated RF signal to be transmitted.
38	X13	Input	N/A	**Crystal Input**. (refer to *Section 4.6*).
35	X13IN	Input	N/A	**Crystal Input**. (refer to *Section 4.6*).
26	X13OUT	Output/Hi-Z	Output	**System Clock**. Buffered 13-MHz system clock.
33	$\overline{PD}$	Input	N/A	**Power Down**. Asserting this input (low), will put the CYWUSB6932/CYWUSB6934 in the Suspend Mode (X13OUT is 0 when $\overline{PD}$ is Low).
14	$\overline{RESET}$	Input	N/A	**Active LOW Reset**. Device reset.
34	PACTL	I/O	Input	**PACTL**. External Power Amplifier control. Pull-down or make output.
20	DIO	I/O	Input	**Data Input/Output.** SERDES Bypass Mode Data Transmit/Receive.
19	DIOVAL	I/O	Input	**Data I/O Valid**. SERDES Bypass Mode Data Transmit/Receive Valid.
21	IRQ	Output /Hi-Z	Output	**IRQ**. Interrupt and SERDES Bypass Mode DIOCLK.
23	MOSI	Input	N/A	**Master-Output-Slave-Input Data**. SPI data input pin.
24	MISO	Output/Hi-Z	Hi-Z	**Master-Input-Slave-Output Data**. SPI data output pin.
25	SCK	Input	N/A	**SPI Input Clock**. SPI clock.
22	$\overline{SS}$	Input	N/A	**Slave Select Enable**. SPI enable.
6, 9, 16, 28, 29, 32, 41, 42, 44, 45	VCC	VCC	H	V_{CC} = 2.7V to 3.6V.
13	GND	GND	L	**Ground = 0V**.
1, 2, 3, 4, 7, 8, 10, 11, 12, 15, 17, 18, 27, 30, 31, 36, 37, 39, 40, 43, 47, 48	NC	N/A	N/A	Must be tied to Ground.
Exposed Paddle	GND	GND	L	Must be tied to Ground.

CYWUSB6934/CYWUSB6932
Top View*



* E-PAD BOTTOMSIDE

** CYWUSB6934 Only

Figure 8-1. CYWUSB6934/CYWUSB6932, 48 QFN – Top View

9.0 Absolute Maximum Ratings

Storage Temperature .....–65°C to +150°C
Ambient Temperature with Power Applied .....–55°C to +125°C
Supply Voltage on V_{CC} relative to VSS.....–0.3V to +3.9V
DC Voltage to Logic Inputs[8] ..... –0.3V to V_{CC} +0.3V
DC Voltage applied to Outputs in High-Z State..... –0.3V to V_{CC} +0.3V
Static Discharge Voltage (Digital)[9] ..... >2000V
Static Discharge Voltage (RF)[9]..... 500V
Latch-up Current.....+200 mA, –200 mA

10.0 Operating Conditions

V_{CC} (Supply Voltage).....2.7V to 3.6V
T_A (Ambient Temperature Under Bias) .....0°C to +70°C
Ground Voltage..... 0V
F_{OSC} (Oscillator or Crystal Frequency)..... 13 MHz ±50 ppm

11.0 DC Characteristics (Over the Operating Range)

Table 11-1. DC Parameters

Parameter	Description	Conditions	Min.	Typ.[11]	Max.	Unit
V_{CC}	Supply Voltage		2.7	3.0	3.6	V
V_{OH1}	Output High Voltage condition 1	At I_{OH} = –100.0 µA	V_{CC} – 0.1	V_{CC}		V
V_{OH2}	Output High Voltage condition 2	At I_{OH} = –2.0 mA	2.4	3.0		V
V_{OL}	Output Low Voltage	At I_{OL} = 2.0 mA		0.0	0.4	V
V_{IH}	Input High Voltage		2.0		V_{CC}[10]	V
V_{IL}	Input Low Voltage		–0.3		0.8	V
I_{IL}	Input Leakage Current	$0 < V_{IN} < V_{CC}$	–1	0.26	+1	µA
C_{IN}	Pin Input Capacitance (except X13, X13IN, RFIN)			3.5	10	pF
I_{Sleep}	Current consumption during power-down mode	$\overline{PD}$ = LOW		0.24	10[14]	µA
IDLE I_{CC}	Current consumption without synthesizer	$\overline{PD}$ = HIGH		3		mA
STARTUP I_{CC}	ICC from $\overline{PD}$ high to oscillator stable.			1.8		mA
TX AVG I_{CC1}	Average transmitter current consumption[12]	no handshake		5.9		mA
TX AVG I_{CC2}	Average transmitter current consumption[13]	with handshaking		8.1		mA
RX $I_{CC (PEAK)}$	Current consumption during receive			57.7		mA
TX $I_{CC (PEAK)}$	Current consumption during transmit			69.1		mA
SYNTH SETTLE I_{CC}	Current consumption with Synthesizer on, No Transmit or Receive			28.7		mA

Notes:
8. It is permissible to connect voltages above Vcc to inputs through a series resistor limiting input current to 1 mA. This can't be done during power down mode. AC timing not guaranteed.
9. Human Body Model (HBM).
10. It is permissible to connect voltages above Vcc to inputs through a series resistor limiting input current to 1 mA.
11. Typ. values measured with V_{CC} = 3.0V @ 25°C
12. Average Icc when transmitting a 5-byte packet (3 data bytes + 2 bytes of protocol) every 10 ms using the WirelessUSB LS 1-way protocol.
13. Average Icc when transmitting a 5-byte packet (3 data bytes + 2 bytes of protocol) every 10 ms using the WirelessUSB LS 2-way protocol.
14. Max. value measured with V_{CC} = 3.3V

12.0 AC Characteristics [15]

Table 12-1. SPI Interface[17]

Parameter	Description	Min.	Typ.	Max.	Unit
t_{SCK_CYC}	SPI Clock Period	476			ns
$t_{SCK\_HI\ (BURST\ READ)}$[16]	SPI Clock High Time	238			ns
t_{SCK_HI}	SPI Clock High Time	158			ns
t_{SCK_LO}	SPI Clock Low Time	158			ns
t_{DAT_SU}	SPI Input Data Set-up Time	10			ns
t_{DAT_HLD}	SPI Input Data Hold Time	97[17]			ns
t_{DAT_VAL}	SPI Output Data Valid Time	77[17]		174[17]	ns
t_{SS_SU}	SPI Slave Select Set-up Time before first positive edge of SCK[18]	250			ns
t_{SS_HLD}	SPI Slave Select Hold Time after last negative edge of SCK	80			ns

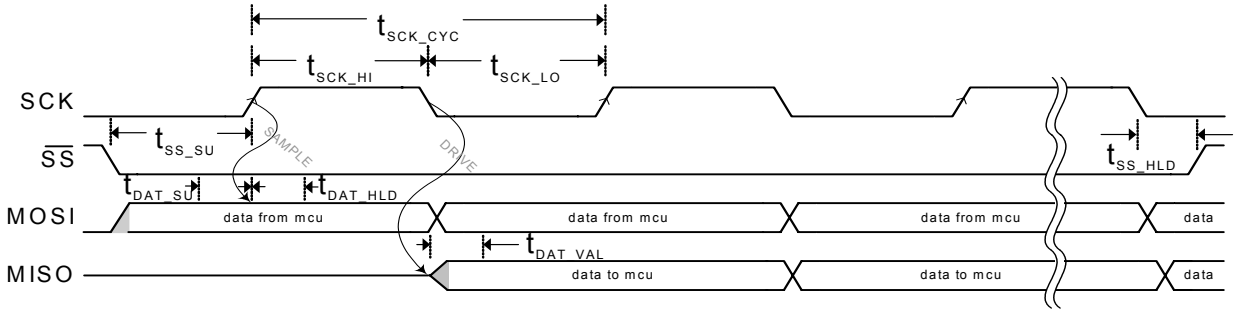


Figure 12-1. SPI Timing Diagram

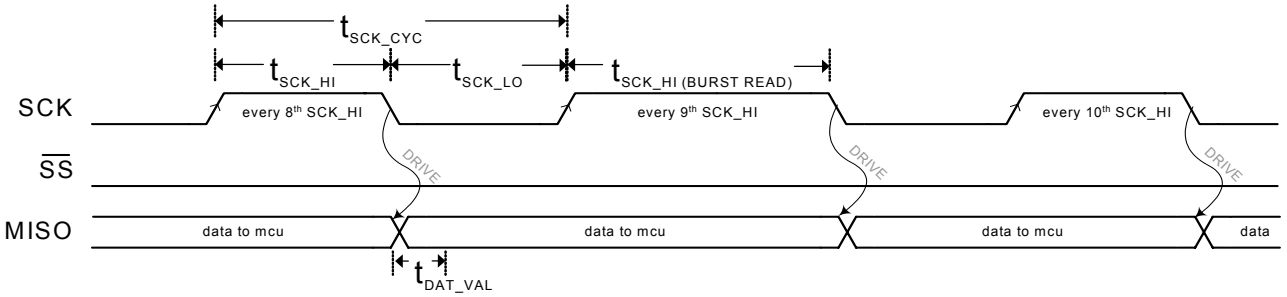


Figure 12-2. SPI Burst Read Every 9th SCK HI Stretch Timing Diagram

Notes:
15. AC values are not guaranteed if voltages on any pin exceed Vcc.
16. This stretch only applies to every 9th SCK HI pulse for SPI Burst Reads only.
17. For F_{OSC} = 13 MHz ±50ppm, 3.3v @ 25°C.
18. SCK must start low, otherwise the success of SPI transactions are not guaranteed.

Table 12-2. DIO Interface

Parameter	Description				
Transmit		**Min.**	**Typ.**	**Max.**	**Unit**
$t_{TX\_DIOVAL\_SU}$	DIOVAL Set-up Time	2.1			µs
$t_{TX\_DIO\_SU}$	DIO Set-up Time	2.1			µs
$t_{TX\_DIOVAL\_HLD}$	DIOVAL Hold Time	0			µs
$t_{TX\_DIO\_HLD}$	DIO Hold Time	0			µs
$t_{TX\_IRQ\_HI}$	Minimum IRQ High Time - 32 chips/bit DDR		8		µs
	Minimum IRQ High Time - 32 chips/bit		16		µs
	Minimum IRQ High Time - 64 chips/bit		32		µs
$t_{TX\_IRQ\_LO}$	Minimum IRQ Low Time - 32 chips/bit DDR		8		µs
	Minimum IRQ Low Time - 32 chips/bit		16		µs
	Minimum IRQ Low Time - 64 chips/bit		32		µs
Receive		**Min.**	**Typ.**	**Max.**	**Unit**
$t_{RX\_DIOVAL\_VLD}$	DIOVAL Valid Time - 32 chips/bit DDR	–0.01		6.1	µs
	DIOVAL Valid Time - 32 chips/bit	–0.01		8.2	µs
	DIOVAL Valid Time - 64 chips/bit	–0.01		16.1	µs
$t_{RX\_DIO\_VLD}$	DIO Valid Time - 32 chips/bit DDR	–0.01		6.1	µs
	DIO Valid Time - 32 chips/bit	–0.01		8.2	µs
	DIO Valid Time - 64 chips/bit	–0.01		16.1	µs
$t_{RX\_IRQ\_HI}$	Minimum IRQ High Time - 32 chips/bit DDR		1		µs
	Minimum IRQ High Time - 32 chips/bit		1		µs
	Minimum IRQ High Time - 64 chips/bit		1		µs
$t_{RX\_IRQ\_LO}$	Minimum IRQ Low Time - 32 chips/bit DDR		8		µs
	Minimum IRQ Low Time - 32 chips/bit		16		µs
	Minimum IRQ Low Time - 64 chips/bit		32		µs

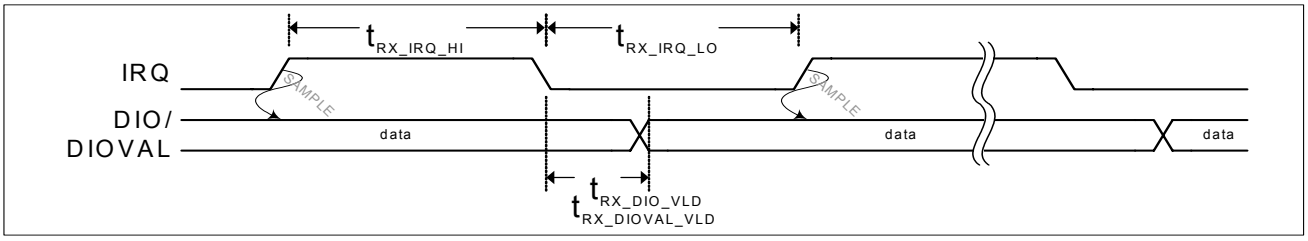


Figure 12-3. DIO Receive Timing Diagram

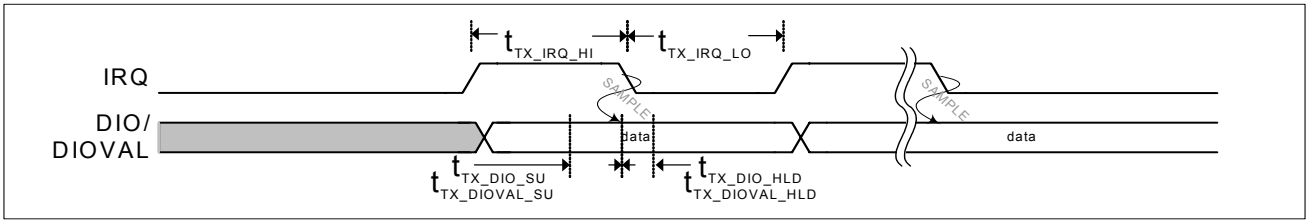


Figure 12-4. DIO Transmit Timing Diagram

12.1 Radio Parameters

Table 12-3. Radio Parameters

Parameter Description	Conditions	Min.	Typ.	Max.	Unit
RF Frequency Range	[20]	2.400		2.483	GHz
Radio Receiver ($T = 25°C$, $V_{CC} = 3.3V$, fosc = 13.000 MHz, X13OUT off, 64 chips/bit, Threshold Low = 8, Threshold High = 56, BER $\leq 10^{-3}$)					
Sensitivity			–90		dBm
Maximum Received Signal		–20	–10		dBm
RSSI value for PWR_{in} > –40 dBm			28–31		
RSSI value for PWR_{in} < –95 dBm			0–10		
Interference Performance					
Co-channel Interference rejection Carrier-to-Interference (C/I)	C = –60 dBm		11		dB
Adjacent (1 MHz) channel selectivity C/I 1 MHz	C = –60 dBm		3		dB
Adjacent (2 MHz) channel selectivity C/I 2 MHz	C = –60 dBm		–30		dB
Adjacent (≥ 3 MHz) channel selectivity C/I ≥ 3 MHz	C = –67 dBm		–40		dB
Image[22] Frequency Interference, C/I Image	C = –67 dBm		–20		dB
Adjacent (1 MHz) interference to in-band image frequency, C/I image ±1 MHz	C = –67 dBm		–25		dB
Out-of-Band Blocking Interference Signal Frequency					
30 MHz – 2399 MHz, except (FO/N & FO/N±1 MHz)[19]	C = –67 dBm		–30		dBm
2498 MHz – 12.75 GHz, except (FO*N & FO*N±1 MHz) [19]	C = –67 dBm		–20		dBm
Intermodulation	C = –64 dBm, Δf = 5,10 MHz		–39		dBm
Spurious Emission					
30 MHz–1 GHz				–57	dBm
1 GHz–12.75 GHz except (4.8 GHz - 5.0 GHz)				–54	dBm
4.8 GHz–5.0 GHz				–40[21]	dBm
Radio Transmitter ($T = 25°C$, $V_{CC} = 3.3V$, fosc = 13.000 MHz)					
Maximum RF Transmit Power	PA = 7		0		dBm
RF Power Control Range			30		dB
RF Power Range Control Step Size	seven steps, monotonic		4.3		dB
Frequency Deviation	PN Code Pattern 10101010		270		kHz
Frequency Deviation	PN Code Pattern 11110000		320		kHz
Zero Crossing Error			±125		ns
Occupied Bandwidth	100-kHz resolution bandwidth, –6 dBc	500			kHz
Initial Frequency Offset			±75		kHz
In-band Spurious					
Second Channel Power (±2 MHz)				–30	dBm
≥ Third Channel Power (≥3 MHz)				–40	dBm
Non-Harmonically Related Spurs					
30 MHz–12.75 GHz				–54	dBm
Harmonic Spurs					
Second Harmonic				–28	dBm
Third Harmonic				–25	dBm
Fourth and Greater Harmonics				–42	dBm

Notes:
19. FO = Tuned Frequency, N = Integer.
20. Subject to regulation.
21. Antenna matching network and antenna will attenuate the output signal at these frequencies to meet regulatory requirements.
22. Image frequency is +4 MHz from desired channel (2 MHz low IF, high side injection).

12.2 Power Management Timing

Table 12-4. Power Management Timing (The values below are dependent upon oscillator network component selection)[27]

Parameter	Description	Conditions	Min.	Typ	Max.	Unit
t_{PDN_X13}	Time from $\overline{PD}$ deassert to X13OUT			2000		µs
t_{SPI_RDY}	Time from oscillator stable to start of SPI transactions		1			µs
t_{PWR_RST}	Power On to $\overline{RESET}$ deasserted	V_{CC} @ 2.7V	1300			µs
t_{RST}	Minimum $\overline{RESET}$ asserted pulse width		1			µs
t_{PWR_PD}	Power On to $\overline{PD}$ deasserted[23]		1300			µs
t_{WAKE}	$\overline{PD}$ deassert to clocks running[24]			2000		µs
t_{PD}	Minimum $\overline{PD}$ asserted pulse width		10			µs
t_{SLEEP}	$\overline{PD}$ assert to low power mode			50		ns
t_{WAKE_INT}	$\overline{PD}$ deassert to IRQ[25] assert (wake interrupt)[26]			2000		µs
t_{STABLE}	$\overline{PD}$ deassert to clock stable	to within ±10 ppm		2100		µs
$t_{STABLE2}$	IRQ assert (wake interrupt) to clock stable	to within ±10 ppm		2100		µs

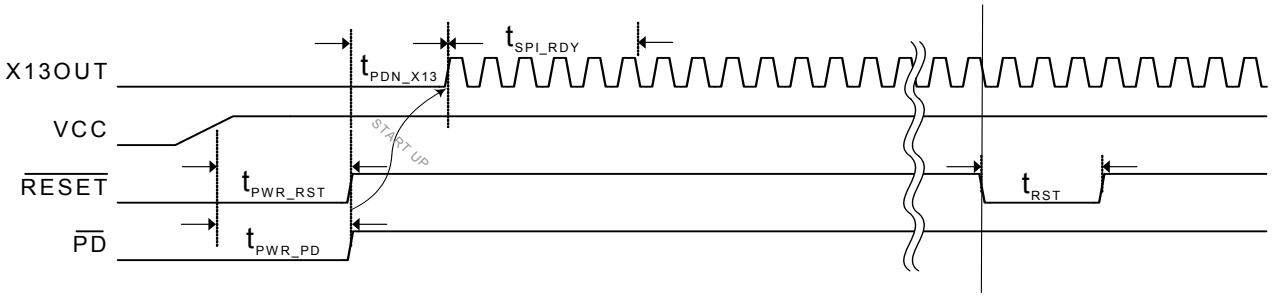


Figure 12-5. Power On Reset/Reset Timing

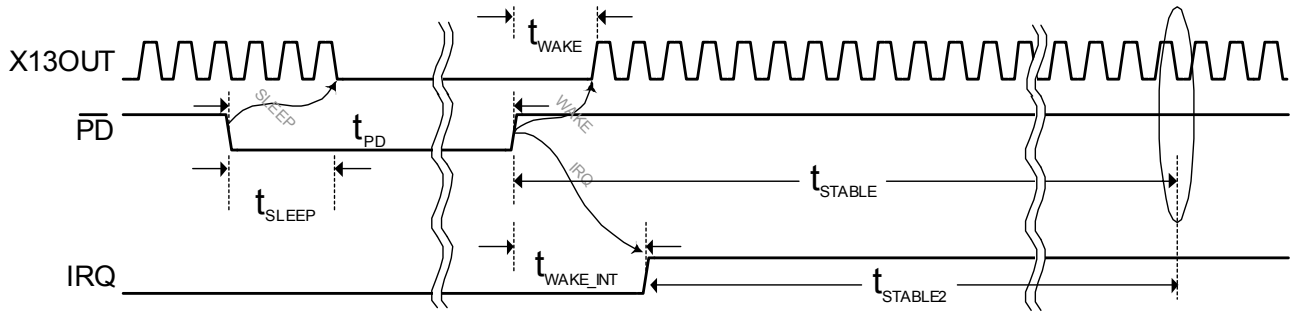
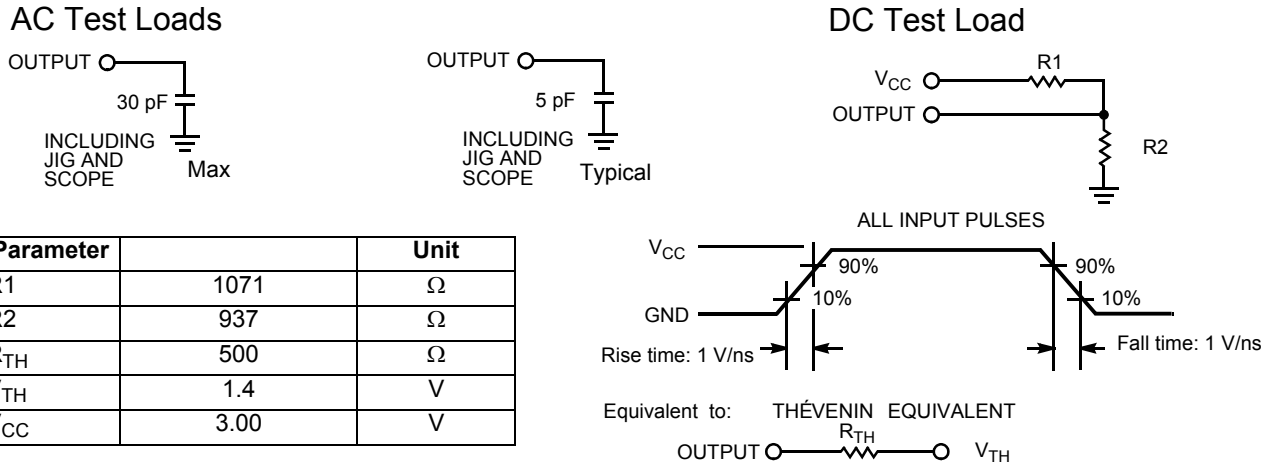


Figure 12-6. Sleep / Wake Timing

Notes:
23. The $\overline{PD}$ pin must be asserted at power up to ensure proper crystal startup.
24. When X13OUT is enabled.
25. Both the polarity and the drive method of the IRQ pin are programmable. See *page 9* for more details. *Figure 12-6* illustrates default values for the Configuration register (Reg 0x05, bits 1:0).
26. A wakeup event is triggered when the $\overline{PD}$ pin is deasserted. *Figure 12-6* illustrates a wakeup event configured to trigger an IRQ pin event via the Wake Enable register (Reg 0x1C, bit 0=1).
27. Measured with CTS ATXN6077A crystal.

12.3 AC Test Loads and Waveforms for Digital Pins



Parameter		Unit
R1	1071	Ω
R2	937	Ω
R_{TH}	500	Ω
V_{TH}	1.4	V
V_{CC}	3.00	V

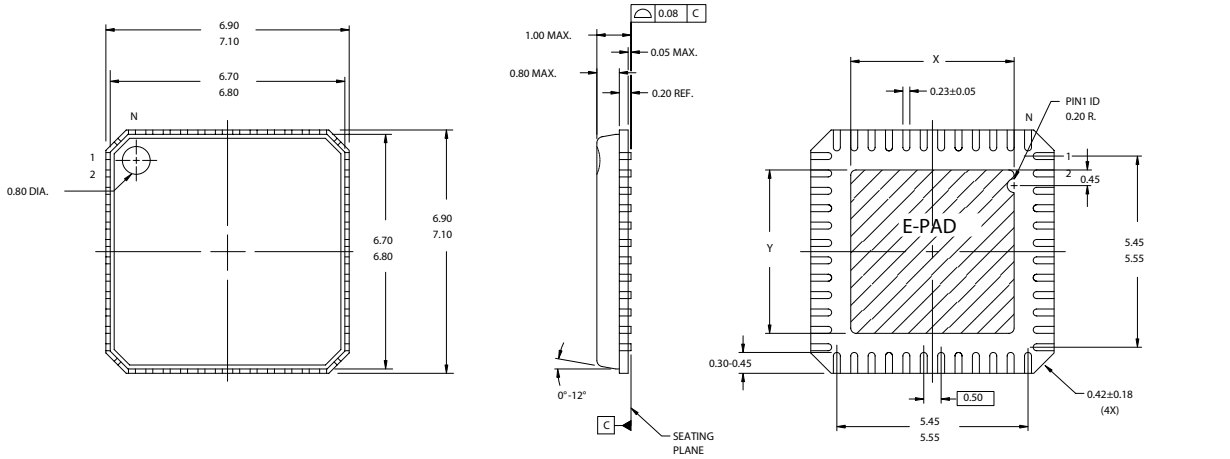
Figure 12-7. AC Test Loads and Waveforms for Digital Pins

13.0 Ordering Information

Table 13-1. Ordering Information

Part Number	Radio	Package Name	Package Type	Operating Range
CYWUSB6932-48LFXC	Transmitter	48 QFN	48 Quad Flat Package No Leads Lead-Free	Commercial
CYWUSB6934-48LFXC	Transceiver	48 QFN	48 Quad Flat Package No Leads Lead-Free	Commercial

14.0 Package Description



DIMENSIONS IN mm MIN.
MAX.

REFERENCE JEDEC MO-220

PKG. WEIGHT 0.13 gms

E-PAD SIZE (X, Y MAX.)	PADDLE SIZE
5.1 X 5.1	5.3 X 5.3
3.8 X 3.8	4.0 X 4.0

Figure 14-1. 48-pin Lead-Free QFN 7 x 7 mm LY48

The recommended dimension of the PCB pad size for the E-PAD underneath the QFN is 209 mils × 209 mils (width x length).

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Document History Page

Document Title: CYWUSB6932/CYWUSB6934 WirelessUSB™ LS 2.4-GHz DSSS Radio SoC Document Number: 38-16007				
REV.	**ECN NO.**	**Issue Date**	**Orig. of Change**	**Description of Change**
**	123907	01/20/03	LXA	New data sheet
*A	125470	04/28/03	XGR	Preliminary release
*B	127076	07/30/03	KKU	Updated pinouts, timing diagrams, AC Test loads, DC Characteristics, Radio Characteristics Removed die
*C	128886	08/04/03	KKV	Minor change: removed table of contents and fixed layout of section 10.
*D	129180	12/04/03	TGE	Updated AC and DC characteristics from char. results Updated register entries Changed package type from 56-pin QFN to 48-pin QFN Updated all pinouts and timing diagrams Updated block diagram and functional description Updated application interfaces Added Interrupt descriptions
*E	131851	12/15/03	TGE	Changed Static Discharge Voltage (Digital) Specification of Section 9.0
*F	241471	See ECN	ZTK	Removed Static Discharge Voltage (Digital) Specification of Section 9.0 footnote Updated REG_DATA_RATE (0x04), 111—Not Valid Swapped bit field descriptions of REG_CONFIG Corrected *Figure 3-1* and *Figure 6-2* Minor edits throughout
*G	284810	See ECN	ZTK	Removed SOIC package option Updated ordering information section Added *Table 4-1* Internal PA Output Power Step Table Added $t_{STABLE2}$ Parameter to *Table 12-4* and *Figure 12-6* Corrected *Figure 14-1* caption Corrected *Figure 6-2* to show QFN matching network Removed Addr 0x01 and 0x02 - unused Updated *Figure 8-1* Updated Spurious Emissions parameters
*H	335758	See ECN	TGE	Corrected *Figure 6-2* - swap RFIN / RFOUT Corrected REG_CONTROL - bit 1 description Added *Table 11-1* footnote 14 - Max. value measured with Vcc = 3.3V