

ENGINEERING STATEMENT

For Type Certification of

MIDLAND USA

Model No: 70-0511C

FCC ID: NCP700511C

I am an Electronics Engineer, a principal in the firm of Hyak Laboratories, Inc., Springfield, Virginia. My education and experience are a matter of record with the Federal Communications Commission.

Hyak Laboratories, Inc. has been authorized by Midland USA to make type certification measurements on the 70-0511C transceiver. These tests made by me or under my supervision in our Springfield laboratory.

Test data and documentation required by the FCC for Type Certification are included in this report. The data verifies that the above mentioned transceiver meets FCC requirements and Type Certification is requested.

Rowland S. Johnson

Dated: October 20, 1999

A. INTRODUCTION

The following data are submitted in connection with this request for Type Certification of the 70-0511C transceiver in accordance with Part 2, Subpart J of the FCC Rules.

The 70-0511C is a VHF, frequency modulated transceiver intended for 25 kHz channel land mobile applications in the 42 - 50 MHz band. It operates from a 13.8 vehicle supply. Output power rating is 60 watts.

B. GENERAL INFORMATION REQUIRED FOR TYPE CERTIFICATION
(Paragraph 2.983 of the Rules)

1. Name of applicant: Midland USA
2. Identification of equipment: NCP700511C
 - a. The equipment identification label is submitted as a separate exhibit.
 - b. Photographs of the equipment are submitted as a separate exhibit.
3. Quantity production is planned.
4. Technical description:
 - a. 16k0F3E emission
 - b. Frequency range: 42 - 50 MHz.
 - c. Operating power of transmitter is fixed at the factory at 60 watts.
 - d. Maximum power permitted under Part 90 of the FCC is 350 watts, and the 70-0511C fully complied with those power limitations.
 - e. The dc voltage and dc currents at final amplifier:

Collector voltage: 13.6 Vdc
Collector current: 9.4 A
 - f. Function of each active semiconductor device:
See Appendix 1.
 - g. Complete circuit diagram is submitted as a separate exhibit.
 - h. A draft instruction book is submitted as a separate exhibit.
 - i. The transmitter tune-up procedure is submitted as a separate exhibit.

B. GENERAL INFORMATION (continued)

- j. A description of circuits for stabilizing frequency is included in Appendix 2.
 - k. A description of circuits and devices employed for suppression of spurious radiation and for limiting modulation is included in Appendix 3.
 - l. Not applicable.
5. Data for 2.985 through 2.997 follow this section.

C. RF POWER OUTPUT (Paragraph 2.985(a) of the Rules)

RF power output was measured with a Bird 4421 RF power meter and a Bird 8325-power attenuator as a 50 ohm dummy load. Maximum power measured was 64 watts.

D. MODULATION CHARACTERISTICS

1. A curve showing frequency response of the transmitter is shown in Figure 1. Reference level was audio signal output from a Boonton 8220 modulation meter with one kHz deviation. Audio output was measured with a Audio Precision System One TRMS voltmeter and tracking generator.
2. Modulation limiting curves are shown in Figure 2 using a Boonton 8220 modulation meter. Signal level was established with a Audio Precision System One TRMS voltmeter. The curves show compliance with paragraphs 2.987(b), and 90.211(c).
3. Figure 3 is a graph of the post-limiter low pass filter which meets the requirements of paragraph 90.211(d)(1) in providing a roll-off of $60\text{Log}f/3$ dB where f is audio frequency in kHz. Measurements were made following EIA RS-152B with an Audio Precision System One selective voltmeter on the Boonton 8220 modulation meter audio output.
4. Occupied Bandwidth
(Paragraphs 2.989(c), 90.209(b)(4) and 90.210(d) of the Rules)

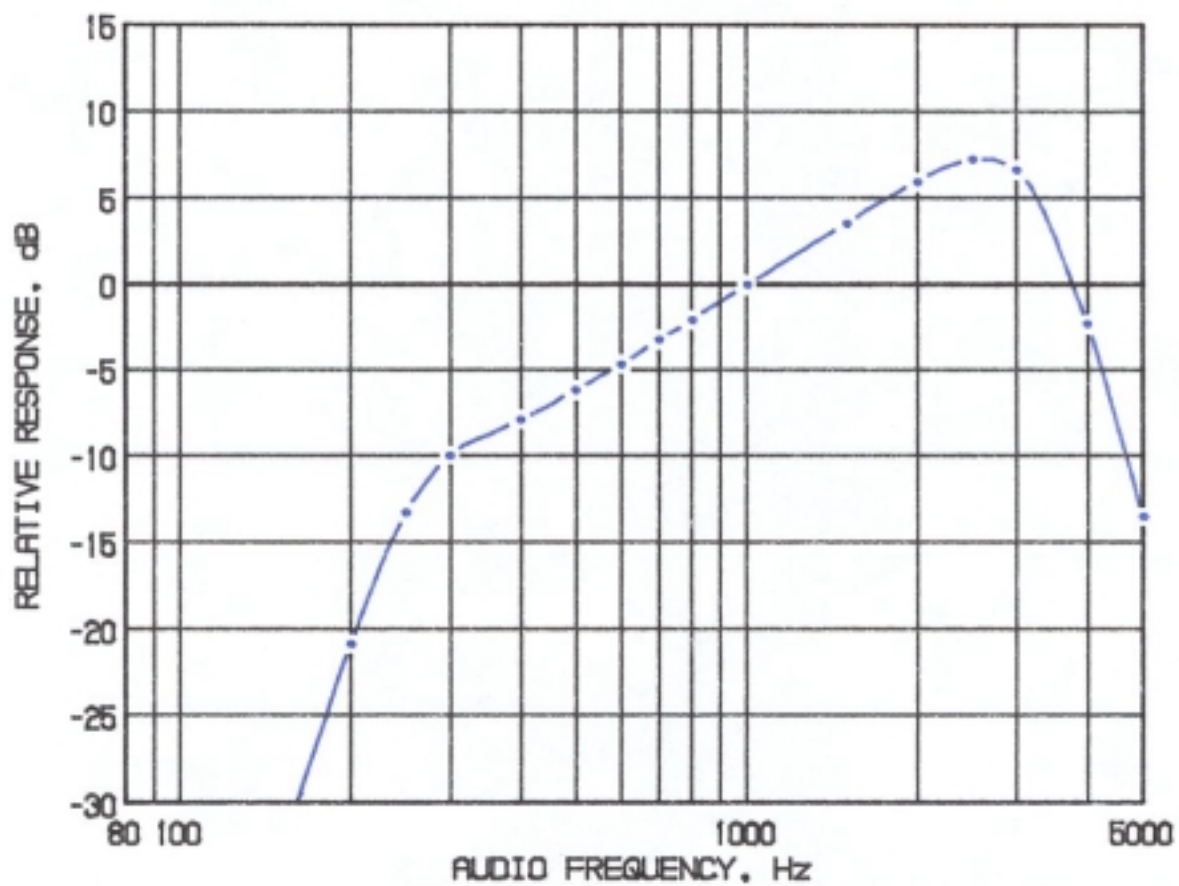
Figure 4 is a plot of the sideband envelope of the transmitter taken with a Advantest R3361A spectrum analyzer. Modulation corresponded to conditions

D. MODULATION CHARACTERISTICS (continued)

of 2.989(c)(1) and consisted of 2500 Hz tone at an input level 16 dB greater than that necessary to produce 50% modulation at 2638 Hz, the frequency of maximum response. Measured modulation under these conditions was 3.9.

The plot has unmodulated carrier as 0 dBm reference.

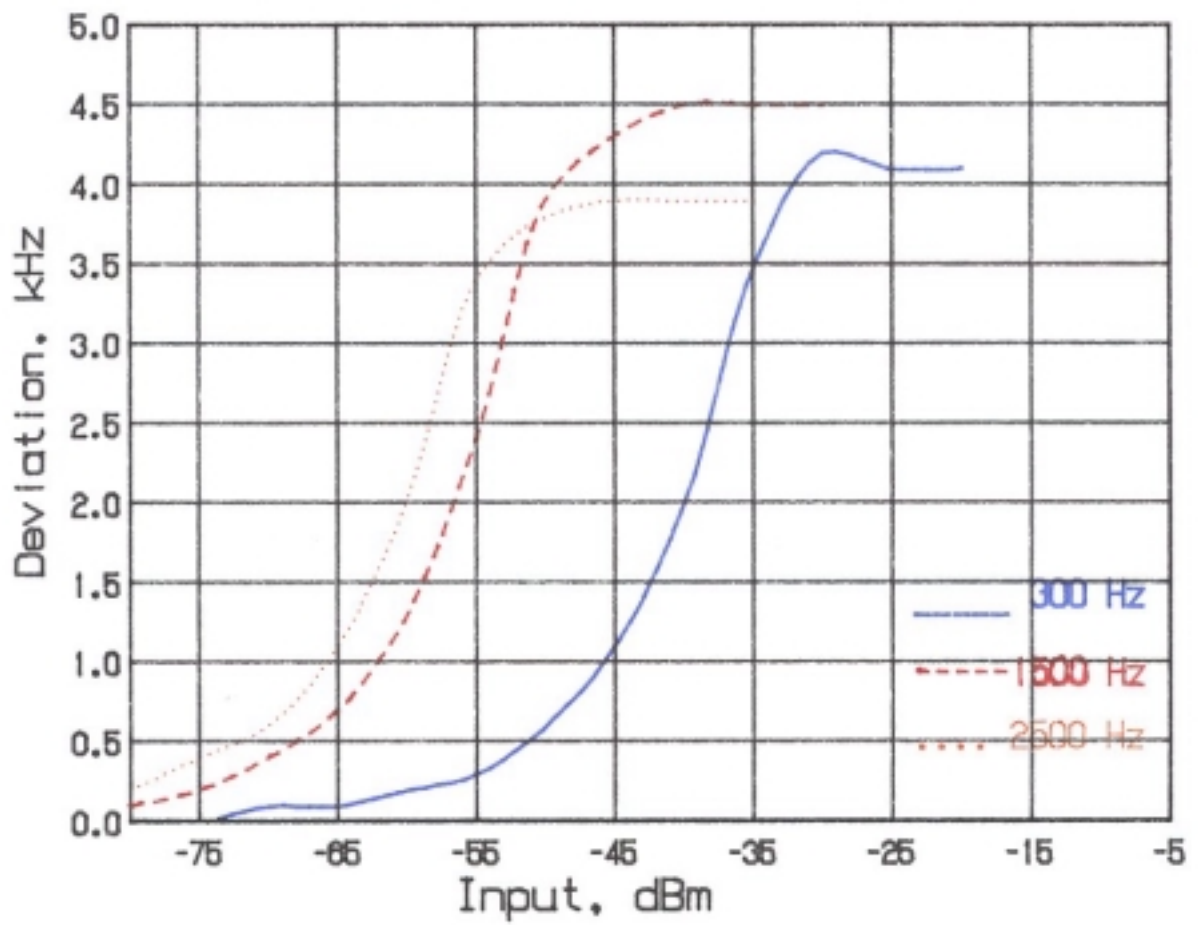
FIGURE 1
MODULATION FREQUENCY RESPONSE



MODULATION FREQUENCY
RESPONSE
FCC ID: NCP700511C

FIGURE 1
FIGURE 2

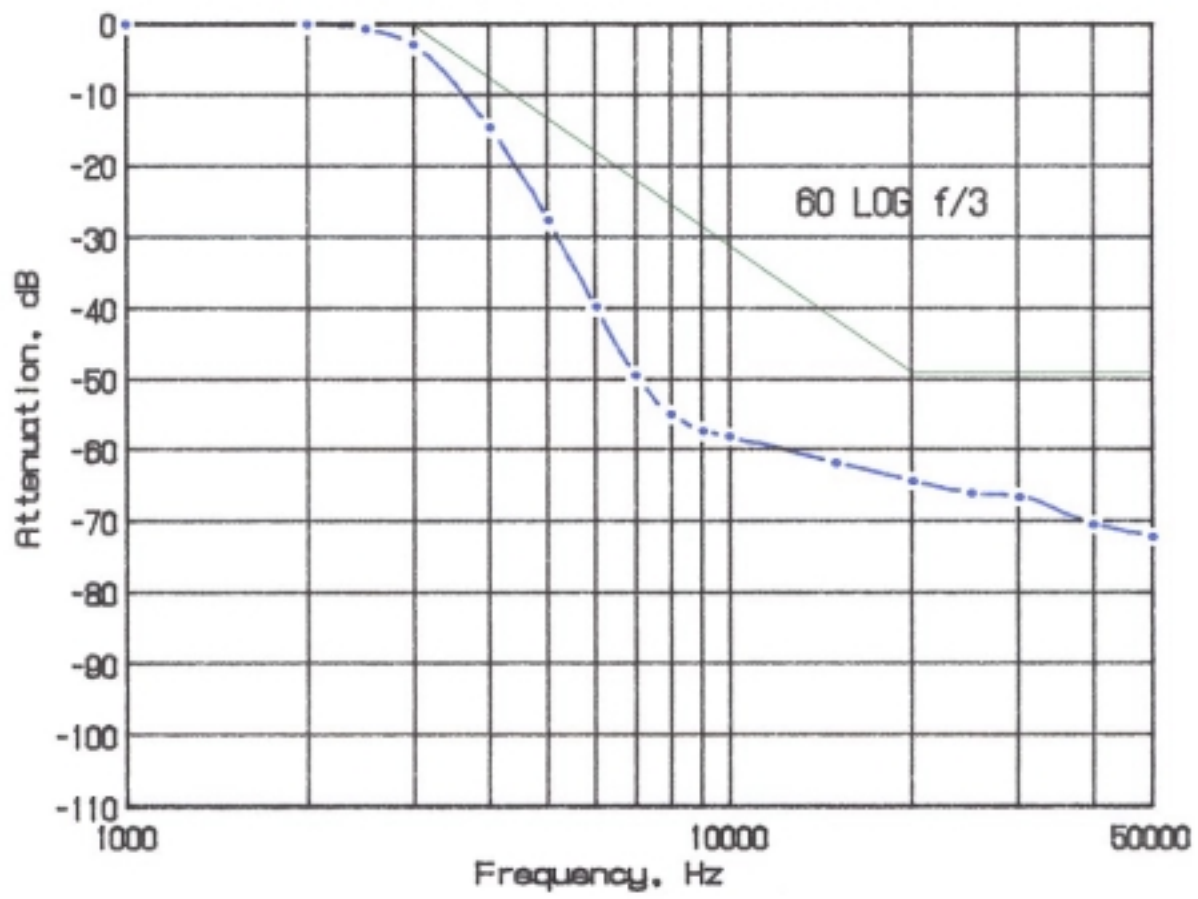
AUDIO LIMITER CHARACTERISTICS



AUDIO LIMITER CHARACTERISTICS
FCC ID: NCP700511C

FIGURE 2
FIGURE 3

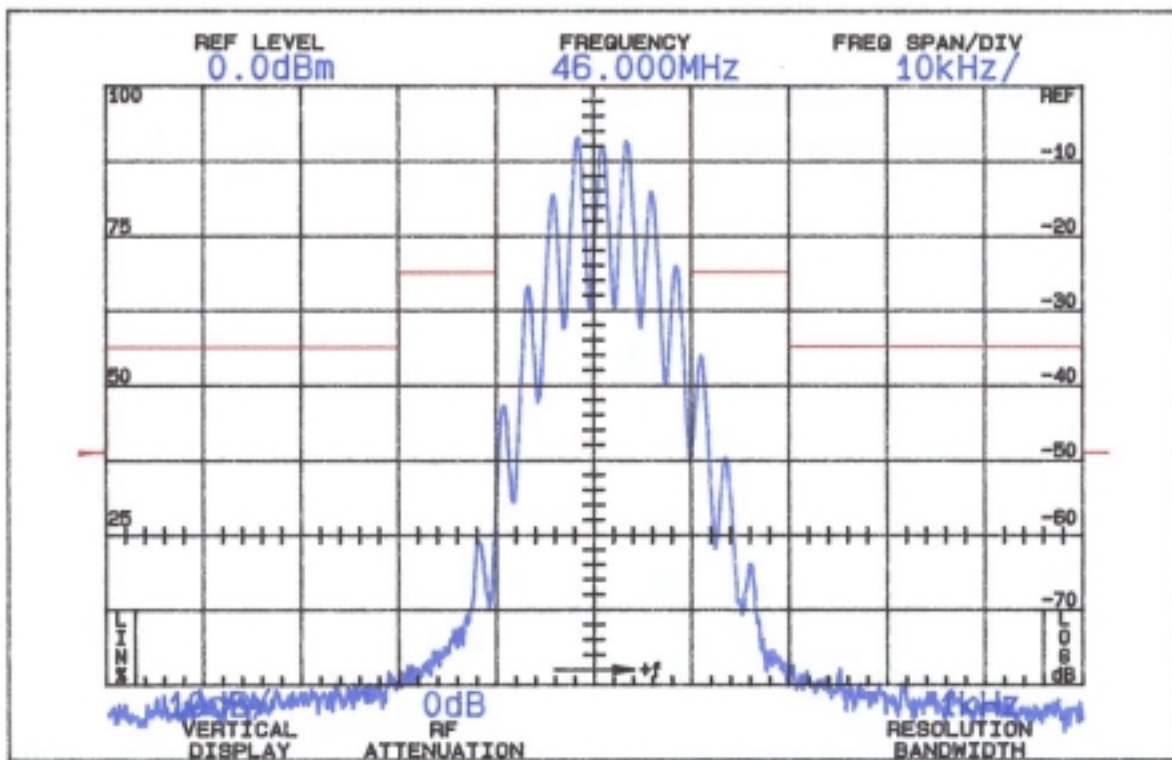
AUDIO LOW PASS FILTER RESPONSE



AUDIO LOW PASS FILTER RESPONSE
FCC ID: NCP700511C

FIGURE 3
FIGURE 4

OCCUPIED BANDWIDTH



ATTENUATION IN dB BELOW
MEAN OUTPUT POWER
Required

On any frequency more than 50%
up to and including 100% of the
authorized bandwidth, 20 kHz
(10-20 kHz)

25

On any frequency more than 100%,
up to and including 250% of the
authorized bandwidth (20-50 kHz)

35

On any frequency removed from
the assigned frequency by more
than 250% of the authorized
bandwidth (over 50 kHz)

$$43 + 10 \log P = 61$$

$$(P = 64 \text{ W})$$

OCCUPIED BANDWIDTH
FCC ID: NCP700511C

FIGURE 4

D. MODULATION CHARACTERISTICS (Continued)

The plot is within the limits imposed by Paragraph 90.211(c) for frequency modulation. The horizontal scale (frequency) is 10 kHz per division and the vertical scale (amplitude) is a logarithmic presentation equal to 10 dB per division.

E. SPURIOUS EMISSIONS AT THE ANTENNA TERMINALS
(Paragraph 2.991 of the Rules)

The 70-0511C transmitter was tested for spurious emissions at the antenna terminals while the equipment was modulated with a 2500 Hz signal, 16 dB above minimum input signal for 50% deviation at 2638 Hz, the frequency of highest sensitivity.

Measurements were made with Tektronix 494P spectrum analyzer coupled to the transmitter output terminal through a Bird 8325 power attenuator. A notch filter was used to attenuate the carrier.

During the tests, the transmitter was terminated in the 50 ohm attenuator. Power was monitored on a Bird 43 Thru-Line wattmeter; dc supply was 13.8 volts throughout the tests.

Spurious emissions were measured at 64 watts output throughout the RF spectrum from 12 (lowest frequency generated in the transmitter is 12.8 MHz) to the tenth harmonic of the carrier.

Any emissions that were between the required attenuation and the noise floor of the spectrum analyzer were recorded. Data are shown in Table 1.

F. DESCRIPTION OF RADIATED SPURIOUS MEASUREMENT FACILITIES

A description of the Hyak Laboratories' radiation test facility is a matter of record with the FCC. The facility meets ANSI 63.4-1992 and was accepted for radiation measurements from 25 to 1000 MHz on October 1, 1976 and is currently listed as an accepted site.

TABLE 1

TRANSMITTER CONDUCTED SPURIOUS
46.000, 13.8 Vdc Input, 64 W

<u>Spurious Frequency MHz</u>	<u>dB Below Carrier Reference</u>
92.000	95
138.000	90

184.000	91
230.000	77
276.000	78
322.000	80
368.000	90
414.000	>106
460.010	>102

Required: $43+10\log(P)$ 61

All other emissions from 12 MHz to the tenth harmonic were 20 dB or more below FCC limit.

*Reference data only, more than 20 dB below FCC limit.

NOTE: Carrier notch filter used to increase dynamic range.

G. FIELD STRENGTH MEASUREMENTS OF SPURIOUS RADIATION

Field intensity measurements of radiated spurious emissions from the Midland 70-0511C were made with a Tektronix 494P spectrum analyzer using Singer DM-105A calibrated dipole antennas below 1 GHz, and Polarad CA-L or CA-S from 1-4.6 GHz.

The transmitter and dummy load were located in an open field 3 meters from the test antenna. Supply voltage was a power supply with a terminal voltage under load of 13.8 Vdc.

Output power was 64 Watts at 46.000 MHz operating frequency. The transmitter and test antennas were arranged to maximize pickup. Both vertical and horizontal test antennae polarization were employed.

Reference level for the spurious radiations was taken as an ideal dipole excited by 64 Watts, the output power of the transmitter according to the following relationship:*

$$E = \frac{(49.2P_t)^{1/2}}{R}$$

where

E = electric-field intensity in volts/meter

P_t = transmitter power in watts

R = distance in meters

for this case $E = \frac{(49.2 \times 64.0)^{1/2}}{3} = 18.7 \text{ V/m}$

Since the spectrum analyzer is calibrated in decibels above one milliwatt (dBm), a conversion, for convenience, was made from dBu to dBm.

$$18.7 \text{ volts/meter} = 18.7 \times 10^6 \text{ uV/m}$$

$$\text{dBu/m} = 20 \text{ Log}_{10}(18.7 \times 10^6)$$

$$= 145 \text{ dBu/m}$$

Since 1 uV/m = -107 dBm, the reference becomes

$$145 - 107 = 38 \text{ dBm}$$

*Reference Data for Radio Engineers, Fourth Edition,
International Telephone and Telegraph Corp., p. 676.

G. FIELD STRENGTH MEASUREMENTS (Continued)

The transmitter and test antennae were arranged to maximize pickup. Both vertical and horizontal test antenna polarization were employed.

The measurement system was capable of detecting signals 95 dB below the reference level. Measurements were made from the lowest frequency generated within the unit (12 MHz), to 10 times operating frequency. Data after application of antenna factors and line loss corrections are shown in Table 2.

TABLE 2

TRANSMITTER CABINET RADIATED SPURIOUS

46.000 MHz, 13.8 Vdc, 64 Watts

<u>Spurious Frequency MHz</u>	<u>dB Below Carrier Reference¹</u>
92.000	74H
138.000	74H

184.000	67H
230.000	73V
276.000	70H
322.000	82*
368.000	76V
414.000	82*
460.000	86*

Required: 61

¹Worst-case polarization, H-Horizontal, V-Vertical.

* Reference data only, more than 20 dB below FCC limit.

All other spurious from 12 MHz to the tenth harmonic were 20 dB or more below FCC limit.

H. FREQUENCY STABILITY (Paragraph 2.995(a)(2) and 90.213 of the Rules)

Measurement of frequency stability versus temperature was made at temperatures from -30°C to +50°C. At each temperature, the unit was exposed to test chamber ambient a minimum of 64 minutes after indicated chamber temperature ambient had stabilized to within $\pm 2^\circ$ of the desired test temperature. Following the 1-hour soak at each temperature, the unit was turned on, keyed and frequency measured within 2 minutes. Test temperature was sequenced in the order shown in Table 3, starting with -30°C.

A Thermotron S1.2 temperature chamber was used. Temperature was monitored with a Keithley 871 digital thermometer. The transmitter output stage was terminated in a dummy load. Primary supply was 13.8 volts. Frequency was measured with a HP 5385A-frequency counter connected to the transmitter through a power attenuator. Measurements were made at 46.000 MHz. No transient keying effects were observed.

TABLE 3
FREQUENCY STABILITY vs. TEMPERATURE
46.000 MHz; 13.8 Vdc; 64 W

<u>Temperature, °C</u>	<u>Output Frequency, Hz</u>	<u>p.p.m.</u>
-29.6	45.999970	-0.7

-19.7	45.999987	-0.3
-10.0	45.999993	-0.2
0.3	46.000007	0.2
10.1	46.000017	0.4
20.0	46.000025	0.5
30.3	46.000027	0.6
39.9	46.000024	0.5
49.7	46.000018	0.4

Maximum frequency error: 45.999970
46.000000

- .000030 MHz

The device met a stability of .002% (20 PPM) or a maximum of \pm .000920 MHz

High Limit 46.000920 MHz
Low Limit 45.999080 MHz

FCC Rule 90.213(a) specifies .002%.

I. FREQUENCY STABILITY AS A FUNCTION OF SUPPLY VOLTAGE
(Paragraph 2.995(d)(2) of the Rules)

Oscillator frequency as a function of power supply voltage was measured with a HP 5385A frequency counter as supply voltage provided by an HP 6264B variable dc power supply was varied from $\pm 15\%$ above the nominal 13.8 volt rating. A Fluke 197 digital voltmeter was used to measure supply voltage at transmitter primary input terminals. Measurements were made at 20°C ambient.

TABLE 4

FREQUENCY STABILITY AS A FUNCTION OF SUPPLY VOLTAGE
46.000 MHz, 13.8 Volts Nominal, 64 W

%	<u>Supply Voltage</u>	<u>Output Frequency, MHz</u>	<u>p.p.m.</u>
115	15.87	46.000021	0.5
110	15.18	46.000021	0.5
105	14.49	46.000022	0.5
100	13.80	46.000025	0.5
95	13.11	46.000029	0.6
90	12.42	46.000030	0.7
85	11.73	46.000030	0.7

Maximum frequency error: 46.000030
46.000000

+ .000030 MHz

The device met a stability of .002% (20 PPM) or a maximum of \pm .000920 MHz.

High Limit	46.000920 MHz
Low Limit	45.999080 MHz

FCC Rule 90.213(a) specifies .002%

APPENDIX 1

FUNCTION OF DEVICES

70-0511 ACTIVE SEMICONDUCTORS (1/3)

SYMBOL	TYPE	FUNCTION
Q201	2SC3357	RF Amp
Q203	2SK508K52	IF Amp
Q204	2SK360E	IF Amp
Q205	RT1N241C	Switch
Q206	RT1N241C	Switch
Q207	IMD3	Switch
IC201	TA31136FN	FM.I.F.
IC202	TDA1519A	Audio Amplifier
IC203	HA178L05UA	5V Regulator
Q301	RT1N241C	Switch
Q302	2SC2462C	Switch
Q303	2SD1421ED	Switch
Q304	RT1N241C	Switch
IC301	HA178L05UA	5V Regulator
IC302	M38203	Micro Computer
IC303	XC61AC4102	Reset Switch
IC305	HD14538	Switch
IC306	BU4066BCFV	Analog Switch

Q401	2SC2462C	Buffer Amplifier
Q402	2SC2462C	Switch
Q403	2SC2462C	Switch
Q404	RT1N241C	Switch
Q405	RT1N241C	Switch
Q406	UMH1N	Switch
Q407	UMH1N	Switch
Q408	RT1N241C	Switch
IC401	AK2368	Audio Processor
IC402	BA3308F	MIC Amplifier
IC403	MF6CWMX	Operational Amplifier
IC404	BA14741F	Operational Amplifier

70-0511 ACTIVE SEMICONDUCTORS (2/3)

SYMBOL	TYPE	FUNCTION
IC405	BA14741F	Operational Amplifier
IC406	HD14053BFP	Analog Switch
IC407	BU4066BCFV	Analog Switch
IC408	BA728F	Buffer Amplifier
IC409	BA728F	Buffer Amplifier
IC410	SC14S66F	Analog Switch
IC411	SC14S66F	Analog Switch
IC412	TC4W53FU	Analog Switch
IC413	TC4W53FU	Analog Switch
IC415	BA728F	Operational Amplifier
Q501	2SC2538	Power Amp
Q502	2SC2539	Power Amp
Q503	2SC2694S	Power Amp
Q504	2SB945Q/P	APC
Q505	2SC2462C	APC
IC501	BA728F	Operational Amplifier
Q701	2SA1121C	Charge Pump
Q702	2SC2462C	Charge Pump
Q703	2SA1121C	Charge Pump
Q704	2SC2462C	Charge Pump
Q705	RT1N241C	Switch
Q706	2SK508K52	VCO
Q707	2SC3356	Pre Amp
Q708	2SC3357	Pre Amp
Q709	2SA1576A	Switch
Q710	UMH1N	Switch
Q712	DTB123YK	Switch
Q713	DTB123YK	Switch
Q714	UMH1N	Switch
Q715	2SB1197K	Switch
Q716	2SD596DV3	Filter
Q717	2SC3357	Local Amp

70-0511 ACTIVE SEMICONDUCTORS (3/3)

SYMBOL	TYPE	FUNCTION
IC701	MB1504PF	Synthesizer
IC702	BU4066BCFV	Analog Switch
IC703	μPC1675G	Operational Amplifier
IC704	μPC1675G	Operational Amplifier
IC705	μPC1675G	Operational Amplifier
Q801	2SK508K52	Switch
Q802	2SK508K52	Switch
Q803	3SK151GR	Buffer Amplifier
Q804	2SC2462C	Buffer Amplifier
Q805	2SA1121C	Multi Vibrator
Q806	2SA1121C	Multi Vibrator
Q807	2SC2462C	Buffer Amplifier
Q808	2SA1121C	Buffer Amplifier
Q809	2SA1121C	Rectifier Amplifier
Q810	2SA1121C	Rectifier Amplifier
Q811	2SC2462C	Rectifier Amplifier
Q812	DTC124EKA	Switch
IC801	MC1350DR2	AGC Amplifier
Q901	2SC2462C	Switch
Q902	RT1N241C	Switch
Q903	2SJ246S	Switch
Q904	RT1N241C	Switch
Q905	RT1N241C	Switch
IC901	M38079EFFP	Micro Computer
IC902	AT24C32N	E2PROM
IC904	NJM7808DLA	8V Regulator
IC905	HA178L05UA	5V Regulator
IC906	RH5VL45AA	5V Regulator
IC907	XC61AC4102MR	Reset Switch
IC910	RH5RL50AA	5V Regulator

FUNCTION OF DEVICES
FCC ID: NCP700511C

APPENDIX 1

APPENDIX 2

CIRCUITS AND DEVICES TO STABILIZE FREQUENCY

A 12.8 MHz TCXO referenced PLL circuit establishes and stabilizes output frequency.

CIRCUITS AND DEVICES TO
STABILIZE FREQUENCY
FCC ID: NCP700511C

APPENDIX 2

APPENDIX 3
CIRCUITS TO SUPPRESS SPURIOUS RADIATION,
LIMIT MODULATION AND CONTROL POWER

RF Power Amplifier

A PC-board stripline is used to match Q501-base terminal to the coax. RF impedance at Q501-collector is transformed by PC-board stripline to the base terminal of driver Q502 and the collector of Q502 is transformed to the base of Q503. RF impedance at the collector of final-stage Q503 is again transformed by PC stripline. L511-L513 and C539-C545 comprise the harmonic filter. R511 serve to drain static and other DC potentials from the antenna.

Modulator

Voice signals from the hand microphone are applied to audio filter of IC401, where frequency response is pre-emphasized and splatter filtered. Gain is such that stronger signals bring IC401 into clipping, which limits modulation. Harmonics above the 3 kHz (wide) or 1.5 kHz (narrow) modulation pass-band are removed in IC401. Modulation signals are then adjusted by IC401 so that modulation at limiting will produce transmitted carrier deviation of ± 5 kHz (wide) or 2.5 kHz (narrow). Output of processed voice signals at IC405 pin 8 is fed to the gain control IC401.

Automatic Power Control

A PC stripline ahead of the harmonic filter, and a thin PC runner adjacent to it, serves as a directional coupler. D502 rectifies a small RF sample that is developed across the thin runner, producing a DC voltage that increases with RF power traveling forward into the antenna. This power level sensing voltage is the inverting input of the comparator IC501 pin 6.

The reference voltage applied to the comparator IC501 pin 5 is fed from IC901 pin 71, of which command is controlled by the programmer in alignment mode. Differential amplifier output drives Q504 which is a current source that feeds primary DC to the collector circuit of predriver Q501. The feedback loop, from the directional coupler to Q504, holds RF output power at a constant level that is determined by IC901.

CIRCUITS TO SUPPRESS SPURIOUS
RADIATION, LIMIT MODULATION-
AND CONTROL POWER
FCC ID: NCP700511C